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## DS4424N+T&R

## Maxim Integrated

Digital to Analog Converters - DAC 4Ch I2C 7-Bit Sink-Source

Any questions, please feel free to contact us. info@kaimte.com



#### **General Description**

The DS4422 and DS4424 contain two or four I<sup>2</sup>C programmable current DACs that are each capable of sinking and sourcing current up to 200µA. Each DAC output has 127 sink and 127 source settings that are programmed using the I<sup>2</sup>C interface. The current DAC outputs power up in a high-impedance state.

#### **Applications**

Power-Supply Adjustment Power-Supply Margining Adjustable Current Sink or Source

#### **Ordering Information**

PART	OUTPUTS	TEMP RANGE	PIN- PACKAGE
<b>DS4422</b> N+	2	-40°C to +85°C	14 TDFN-EP
DS4422N+T&R	2	-40°C to +85°C	14 TDFN-EP
DS4424N+	4	-40°C to +85°C	14 TDFN-EP
DS4424N+T&R	4	-40°C to +85°C	14 TDFN-EP

+Denotes a lead(Pb)-free/RoHS-compliant package.

T&R = Tape and reel.

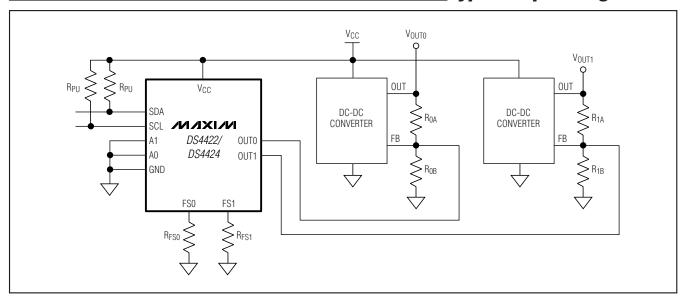
EP = Exposed pad.

#### **Features**

- ♦ Two (DS4422) or Four (DS4424) Current DACs
- ♦ Full-Scale Current 50µA to 200µA
- ♦ Full-Scale Range for Each DAC Determined by **External Resistors**
- ♦ 127 Settings Each for Sink and Source Modes
- ♦ I<sup>2</sup>C-Compatible Serial Interface
- ♦ Two Address Pins Allow Four Devices on Same I<sup>2</sup>C Bus
- **♦ Low Cost**
- ♦ Small Package (14-Pin, 3mm x 3mm TDFN)
- ♦ -40°C to +85°C Temperature Range
- ♦ 2.7V to 5.5V Operating Range

Pin Configuration appears at end of data sheet.

#### Typical Operating Circuit



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#### **ABSOLUTE MAXIMUM RATINGS**

Voltage Range on V <sub>CC</sub> , SDA, and SCL	Operating Temperature Range	40°C to +85°C
Relative to Ground0.5V to +6.0V	Storage Temperature Range	55°C to +125°C
Voltage Range on A0, A1, FS0, FS1, FS2, FS3,	Soldering Temperature	Refer to the IPC/JEDEC
OUTO, OUT1, OUT2, and OUT3 Relative to		J-STD-020 Specification.
Ground0.5V to (V <sub>CC</sub> + 0.5V) (Not to exceed 6.0V.)		

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### RECOMMENDED OPERATING CONDITIONS

 $(T_A = -40^{\circ}C \text{ to } +85^{\circ}C.)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP MAX	UNITS
Supply Voltage	Vcc	(Note 1)	2.7	5.5	V
Input Logic 1 (SDA, SCL, A0, A1)	VIH		0.7 x V <sub>CC</sub>	V <sub>CC</sub> + 0.3	V
Input Logic 0 (SDA, SCL, A0, A1)	VIL		-0.3	0.3 x V <sub>CC</sub>	V
Full-Scale Resistor Values	R <sub>FS0</sub> , R <sub>FS1</sub> , R <sub>FS2</sub> , R <sub>FS3</sub>	(Note 2)	40	160	kΩ

#### DC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +2.7V \text{ to } +5.5V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C.)$ 

PARAMETER	SYMBOL		CONDITIONS	МІМ	l TYP	MAX	UNITS
Supply Current	loo	V <sub>C</sub> C = 5.5V	DS4422			250	
Supply Current	Icc	(Note 3)	DS4424			250	μΑ
Input Leakage (SDA, SCL)	lıL	V <sub>C</sub> C = 5.5V				1	μΑ
Output Leakage (SDA)	IL					1	μΑ
Output Current Low (SDA)	la	$V_{OL} = 0.4V$		3			mA
Output Current Low (SDA)	loL	V <sub>OL</sub> = 0.6V		6			1 IIIA
RFS Voltage	VRFS				0.976	3	V
I/O Capacitance	C <sub>I/O</sub>					10	pF

#### **OUTPUT CURRENT SOURCE CHARACTERISTICS**

 $(V_{CC} = +2.7V \text{ to } +5.5V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C.)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Output Voltage for Sinking Current	Vout:sink	(Note 4)	0.5		3.5	V
Output Voltage for Sourcing Current	Vout:source	(Note 4)	0		V <sub>CC</sub> - 0.75	V
Full-Scale Sink Output Current	IOUT:SINK	(Notes 1, 4)	50		200	μΑ
Full-Scale Source Output Current	IOUT:SOURCE	(Notes 1, 4)	-200		-50	μΑ
Output Current Full-Scale Accuracy	I <sub>OUT:FS</sub>	+25°C, V <sub>CC</sub> = 3.3V; using 0.1% R <sub>FS</sub> resistor (Note 2), V <sub>OUT0</sub> = V <sub>OUT1</sub> = 1.2V			±6	%
Output Current Temperature Coefficient	IOUT:TC	(Note 5)		±75		ppm/°C

#### **OUTPUT CURRENT SOURCE CHARACTERISTICS (continued)**

 $(V_{CC} = +2.7V \text{ to } +5.5V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C.)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Output Current Variation Due to		DC source		0.32		%N
Power-Supply Change		DC sink		0.42		/0/ V
Output Current Variation Due to		DC source, V <sub>OUT</sub> measure at 1.2V		0.16		%N
Output-Voltage Change		DC sink, V <sub>OUT</sub> measure at 1.2V	0.16			/0/V
Output Leakage Current at Zero Current Setting	I <sub>ZERO</sub>		-1		+1	μΑ
Output Current Differential Linearity	DNL	(Notes 6, 7)	-0.5		+0.5	LSB
Output Current Integral Linearity	INL	(Notes 7, 8)	-1		+1	LSB

#### **AC ELECTRICAL CHARACTERISTICS**

 $(V_{CC} = +2.7V \text{ to } +5.5V, T_A = -40^{\circ}C \text{ to } +85^{\circ}C.)$ 

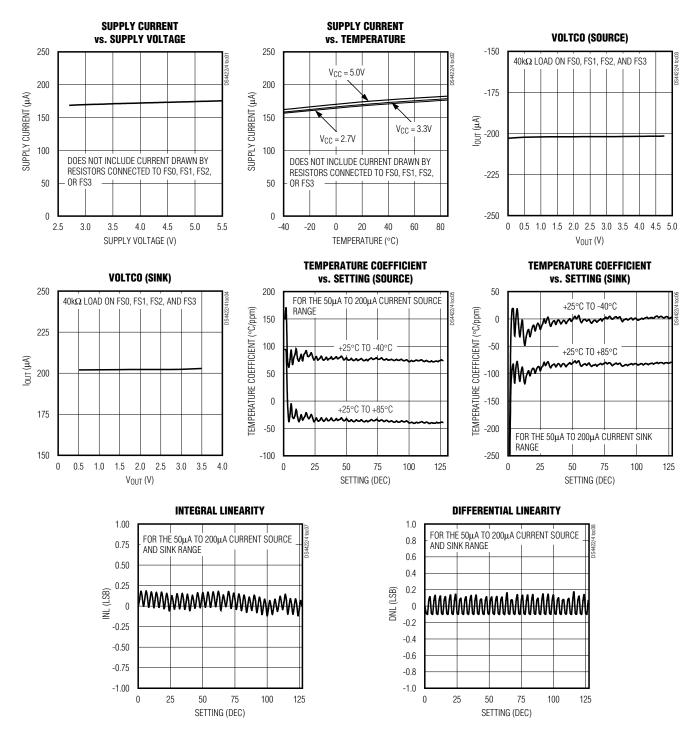
PARAMETER	SYMBOL	CONDITIONS	MIN TYP	MAX	UNITS
SCL Clock Frequency	fscl	(Note 9)	0	400	kHz
Bus Free Time Between STOP and START Conditions	tBUF		1.3		μs
Hold Time (Repeated) START Condition	tHD:STA		0.6		μs
Low Period of SCL	tLOW		1.3		μs
High Period of SCL	tHIGH		0.6		μs
Data Hold Time	tDH:DAT		0	0.9	μs
Data Setup Time	tsu:DAT		100		ns
START Setup Time	tsu:sta		0.6		μs
SDA and SCL Rise Time	t <sub>R</sub>	(Note 10)	20 + 0.1C <sub>B</sub>	300	ns
SDA and SCL Fall Time	tF	(Note 10)	20 + 0.1C <sub>B</sub>	300	ns
STOP Setup Time	tsu:sto		0.6		μs
SDA and SCL Capacitive Loading	СВ	(Note 10)		400	pF

- Note 1: All voltages with respect to ground. Currents entering the IC are specified positive, and currents exiting the IC are negative.
- Note 2: Input resistors (RFS) must be between the specified values to ensure the device meets its accuracy and linearity specifications.
- Note 3: Supply current specified with all outputs set to zero current setting. A0 and A1 are connected to GND. SDA and SCL are connected to V<sub>CC</sub>. Excludes current through R<sub>FS</sub> resistors (I<sub>RFS</sub>). Total current including I<sub>RFS</sub> is I<sub>CC</sub> + (2 x I<sub>RFS</sub>).
- Note 4: The output-voltage range must be satisfied to ensure the device meets its accuracy and linearity specifications.
- **Note 5:** Temperature drift excludes drift caused by external resistor.
- **Note 6:** Differential linearity is defined as the difference between the expected incremental current increase with respect to position and the actual increase. The expected incremental increase is the full-scale range divided by 127.
- Note 7: Guaranteed by design.
- **Note 8:** Integral linearity is defined as the difference between the expected value as a function of the setting and the actual value. The expected value is a straight line between the zero and the full-scale values proportional to the setting.
- Note 9: Timing shown is for fast-mode (400kHz) operation. This device is also backward compatible with I2C standard-mode timing.
- Note 10: C<sub>B</sub>—total capacitance of one bus line in pF.



#### Typical Operating Characteristics

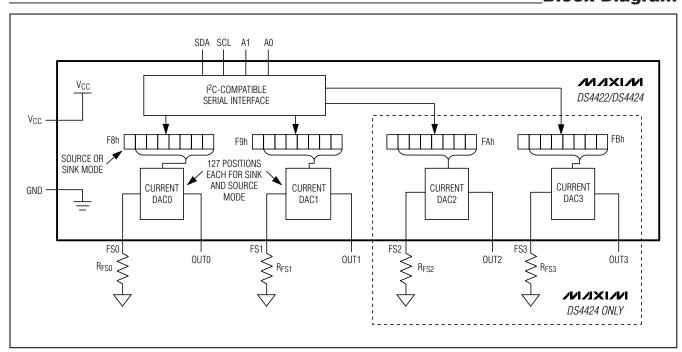
 $(T_A = +25^{\circ}C, \text{ unless otherwise noted.})$ 



#### **Pin Description**

PIN		NAME	FUNCTION				
DS4424	DS4422	NAME	FUNCTION				
1	1	SDA	I <sup>2</sup> C Serial Data. Input/output for I <sup>2</sup> C data.				
2	2	SCL	I <sup>2</sup> C Serial Clock. Input for I <sup>2</sup> C clock.				
3	3	GND	Ground				
4	_	FS3					
5	_	FS2	Full-Scale Calibration Input. A resistor to ground on these pins determines the full-scale				
6	6	FS1	current for each output. FS0 controls OUT0, FS1 controls OUT1, etc. (The DS4422 has only two inputs: FS0 and FS1.)				
7	7	FS0					
8	8	OUT0					
10	10	OUT1	Current Output. Sinks or sources the current determined by the I <sup>2</sup> C interface and the				
12	_	OUT2	resistance connected to FSx. (The DS4422 has only two outputs: OUT0 and OUT1.)				
14	_	OUT3					
9, 11	9, 11	A0, A1	Address Select Inputs. Determines the $I^2C$ slave address by connecting $V_{CC}$ or GND. See the <i>Detailed Description</i> section for the available device addresses.				
13	13	Vcc	Power Supply				
_	4, 5, 12, 14	N.C.	No Connection				
_	_	EP	Exposed Pad. Connect to GND or leave unconnected.				

#### **Block Diagram**



#### **Detailed Description**

The DS4422/DS4424 contain two or four I<sup>2</sup>C adjustable current sources that are each capable of sinking and sourcing current. Each output (OUT0, OUT1, OUT2, and OUT3) has 127 sink and 127 source settings that can be controlled by the I<sup>2</sup>C interface. The full-scale ranges and corresponding step sizes of the outputs are determined by external resistors, connected to pins FS0, FS1, FS2, and FS3, that can adjust the output current over a 4:1 range. Pins OUT2, OUT3, FS2, and FS3 are only available on the DS4424.

The formula to determine RFS (connected to the FSx pins) to attain the desired full-scale current range is:

#### Equation 1:

$$R_{FS} = \frac{V_{RFS}}{16 \times I_{ES}} \times 127$$

Where IFS is the desired full-scale current value, VRFS is the RFS voltage (see the *DC Electrical Characteristics* table), and RFS is the external resistor value.

To calculate the output current value (IOUT) based on the corresponding DAC value (see Table 1 for corresponding memory addresses), use equation 2.

#### **Equation 2:**

$$I_{OUT} = \frac{DAC \ Value(dec)}{127} \times I_{FS}$$

On power-up the DS4422/DS4424 output zero current. This is done to prevent them from sinking or sourcing an incorrect amount of current before the system host controller has had a chance to modify the device's setting.

As a source for biasing instrumentation or other circuits, the DS4422/DS4424 provide a simple and inexpensive current source with an I<sup>2</sup>C interface for control. The adjustable full-scale range allows the application to get the most out of its 7-bit sink or source resolution.

When used in adjustable power-supply applications (see *Typical Operating Circuit*), the DS4422/DS4424 do not affect the initial power-up voltage of the supply because they default to providing zero output current on power-up. As the devices source or sink current into the feedback-voltage node, they change the amount of output voltage required by the regulator to reach its steady-state operating point. Using the external resistor, RFS, to set the output current range, the DS4422/DS4424 provide some flexibility for adjusting the impedances of the feedback network or the range over which the power supply can be controlled or margined.

#### I<sup>2</sup>C Slave Address

The DS4422/DS4424 respond to one of four I $^2$ C slave addresses determined by the two address inputs, A0 and A1. The address inputs should be connected to either V $_{\rm CC}$  or ground. Table 1 lists the slave addresses determined by the address input combinations.

**Table 1. Slave Addresses** 

A1	Α0	SLAVE ADDRESS (HEX)
GND	GND	20h
GND	Vcc	60h
Vcc	GND	A0h
Vcc	Vcc	E0h

#### **Memory Organization**

To control the DS4422/DS4424's current sources, write to the memory addresses listed in Table 2.

**Table 2. Memory Addresses** 

MEMORY ADDRESS (HEX)	CURRENT SOURCE
F8h	OUT0
F9h	OUT1
FAh*	OUT2*
FBh*	OUT3*

<sup>\*</sup>Only for DS4424.

The format of each output control register is given by:

MS	В							LSB
S		D <sub>6</sub>	D <sub>5</sub>	D <sub>4</sub>	D <sub>3</sub>	D <sub>2</sub>	D <sub>1</sub>	D <sub>0</sub>

#### Where:

BIT	NAME	FUNCTION	POWER-ON DEFAULT
S	Sign Bit	Determines if DAC sources or sinks current. For sink S = 0; for source S = 1.	0b
D <sub>X</sub>	Data	7-Bit Data Controlling DAC Output. Setting 0000000b outputs zero current regardless of the state of the sign bit.	0000000b

Example:  $R_{FSO} = 80k\Omega$  and register 0xF8h is written to a value of 0xAAh. Calculate the output current.

 $I_{FS} = (0.976 \text{V}/80 \text{k}\Omega) \times (127/16) = 96.838 \mu\text{A}$ 

The MSB of the output register is 1, so the output is sourcing the value corresponding to position 2Ah (42 decimal). The magnitude of the output current is equal to:

 $96.838\mu A \times (42/127) = 32.025\mu A$ 

## I<sup>2</sup>C Serial Interface Description I<sup>2</sup>C Definitions

The following terminology is commonly used to describe I<sup>2</sup>C data transfers:

**I<sup>2</sup>C Slave Address:** The slave address of the DS4422/DS4424 is determined by the state of the A0 and A1 pins (see Table 1).

**Master Device:** The master device controls the slave devices on the bus. The master device generates SCL clock pulses and START and STOP conditions.

**Slave Devices:** Slave devices send and receive data at the master's request.

**Bus Idle or Not Busy:** Time between STOP and START conditions when both SDA and SCL are inactive and in their logic-high states. When the bus is idle it often initiates a low-power mode for slave devices.

**START Condition:** A START condition is generated by the master to initiate a new data transfer with a slave. Transitioning SDA from high to low while SCL remains high generates a START condition. See Figure 1 for applicable timing.

**STOP Condition:** A STOP condition is generated by the master to end a data transfer with a slave. Transitioning SDA from low to high while SCL remains high generates a STOP condition. See Figure 1 for applicable timing.

Repeated START Condition: The master can use a repeated START condition at the end of one data transfer to indicate that it will immediately initiate a new data transfer following the current one. Repeated STARTs are commonly used during read operations to identify a specific memory address to begin a data transfer. A repeated START condition is issued identically to a normal START condition. See Figure 1 for applicable timing.

**Bit Write:** Transitions of SDA must occur during the low state of SCL. The data on SDA must remain valid and unchanged during the entire high pulse of SCL, plus the setup and hold time requirements (Figure 1). Data is shifted into the device during the rising edge of the SCL.

**Bit Read:** At the end of a write operation, the master must release the SDA bus line for the proper amount of setup time (Figure 1) before the next rising edge of SCL during a bit read. The device shifts out each bit of data on SDA at the falling edge of the previous SCL pulse and the data bit is valid at the rising edge of the current SCL pulse. Remember that the master generates all SCL clock pulses, including when it is reading bits from the slave.

**Acknowledgement (ACK and NACK):** An Acknowledgement (ACK) or Not Acknowledge (NACK) is always the ninth bit transmitted during a byte transfer. The device receiving data (the master during a read or the slave during a write operation) performs an ACK by transmitting a zero during the ninth bit. A device performs a

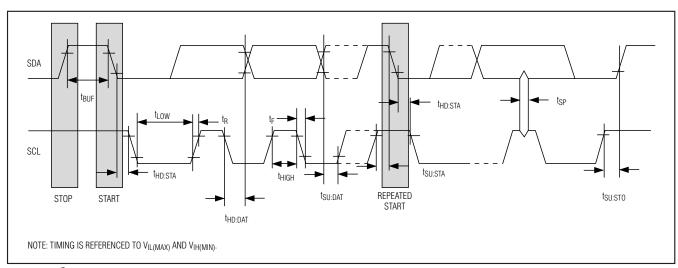


Figure 1. I<sup>2</sup>C Timing Diagram

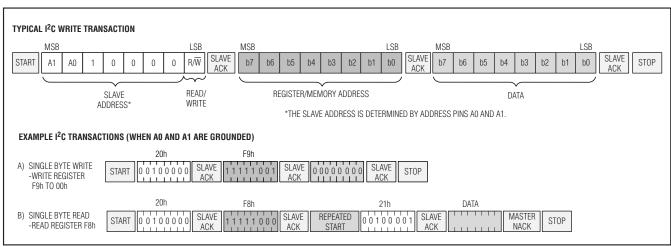


Figure 2. I<sup>2</sup>C Communication Examples

NACK by transmitting a one during the ninth bit. Timing for the ACK and NACK is identical to all other bit writes (Figure 2). An ACK is the acknowledgment that the device is properly receiving data. A NACK is used to terminate a read sequence or as an indication that the device is not receiving data.

**Byte Write:** A byte write consists of 8 bits of information transferred from the master to the slave (most significant bit first) plus a 1-bit acknowledgement from the slave to the master. The 8 bits transmitted by the master are done according to the bit-write definition, and the acknowledgement is read using the bit-read definition.

**Byte Read:** A byte read is an 8-bit information transfer from the slave to the master plus a 1-bit ACK or NACK from the master to the slave. The 8 bits of information that are transferred (most significant bit first) from the slave to the master are read by the master using the bit-read definition above, and the master transmits an ACK using the bit write definition to receive additional data bytes. The master must NACK the last byte read to terminated communication so the slave will return control of SDA to the master.

**Slave Address Byte:** Each slave on the  $I^2C$  bus responds to a slave address byte sent immediately following a START condition. The slave address byte contains the slave address in the most significant 7 bits and the  $R/\overline{W}$  bit in the least significant bit. The DS4422/DS4424 slave address is determined by the

state of the A0 and A1 address pins. Table 1 describes the addresses corresponding to the state of A0 and A1.

When the  $R / \overline{W}$  bit is 0 (such as in A0h), the master is indicating that it will write data to the slave. If  $R / \overline{W} = 1$  (A1h in this case), the master is indicating that it wants to read from the slave. If an incorrect slave address is written, the DS4422/DS4424 assume the master is communicating with another I²C device and ignore the communication until the next START condition is sent.

**Memory Address:** During an I<sup>2</sup>C write operation, the master must transmit a memory address to identify the memory location where the slave is to store the data. The memory address is always the second byte transmitted during a write operation following the slave address byte.

#### I<sup>2</sup>C Communication

Writing to a Slave: The master must generate a START condition, write the slave address byte ( $R/\overline{W}=0$ ), write the memory address, write the byte of data, and generate a STOP condition. Remember that the master must read the slave's acknowledgement during all byte-write operations.

**Reading from a Slave:** To read from the slave, the master generates a START condition, writes the slave address byte with  $R/\overline{W}=1$ , reads the data byte with a NACK to indicate the end of the transfer, and generates a STOP condition.

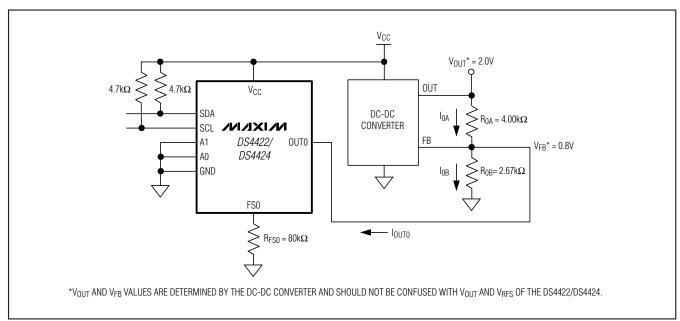


Figure 3. Example Application Circuit

#### **Applications Information**

## Example Calculations for an Adjustable Power Supply

In this example, the *Typical Operating Circuit* is used as a base to create Figure 3, a DC-DC output voltage of 2.0V with  $\pm 20\%$  margin. The adjustable power supply has a DC-DC converter output voltage, V<sub>OUT</sub>, of 2.0V and a DC-DC converter feedback voltage, V<sub>FB</sub>, of 0.8V. To determine the relationship of R<sub>0A</sub> and R<sub>0B</sub>, start with the equation:

$$V_{FB} = \frac{R_{0B}}{R_{0A} + R_{0B}} \times V_{OUT}$$

Substituting  $V_{FB} = 0.8V$  and  $V_{OUT} = 2.0V$ , the relationship between  $R_{OA}$  and  $R_{OB}$  is determined to be:

$$R_{0A} \approx 1.5 \times R_{0B}$$

 $I_{OUTO}$  is chosen to be  $100\mu A$  (midrange source/sink current for the DS4422/DS4424). Summing the currents into the feedback node produces the following:

Where:

$$I_{ROB} = \frac{V_{FB}}{R_{OB}}$$

And:

$$I_{ROA} = \frac{V_{OUT} - V_{FB}}{R_{OA}}$$

To create a 20% margin in the supply voltage, the value of VOUT is set to 2.4V. With these values in place,  $R_{0B}$  is calculated to be 2.67k $\Omega$ , and  $R_{0A}$  is calculated to be 4.00k $\Omega$ . The current DAC in this configuration allows the output voltage to be moved linearly from 1.6V to 2.4V using 127 settings. This corresponds to a resolution of 6.3mV/step.

#### **Vcc Decoupling**

To achieve the best results when using the DS4422/DS4424, decouple the power supply with a 0.01µF or 0.1µF capacitor. Use a high-quality ceramic surfacemount capacitor if possible. Surface-mount components minimize lead inductance, which improves performance, and ceramic capacitors tend to have adequate high-frequency response for decoupling applications.

#### **Power Rail Considerations**

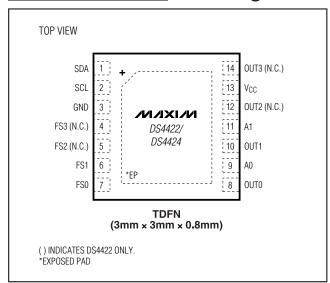
Given that the absolute maximum rating for the OUT pins is  $V_{CC}$  + 0.5V, it is recommended that the DS4424 power rail be brought up before or at the same time as the power rail of the source it is controlling.

### Pin Configuration

#### Package Information

For the latest package outline information, go to <a href="https://www.maxim-ic.com/packages">www.maxim-ic.com/packages</a>.

PACKAGE TYPE	PACKAGE CODE	DOCUMENT NO.	
14 TDFN-EP	T1433+2	<u>21-0137</u>	



#### **Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	3/08	Initial release.	_
1	7/09	Added the Power Rail Considerations section.	9

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