
maXTouch 640-node Touchscreen Controller

DATASHEET**Features**

- Atmel® maXTouch® Adaptive Sensing Touchscreen Technology
 - Up to 32 X (transmit) lines and 20 Y (receive) lines
 - A maximum of 640 nodes can be allocated to the touchscreen
 - Screen sizes of 6.2 inches diagonal are supported while meeting Microsoft® Phone touch performance requirements
 - Multi-touch support with up to 16 concurrent touches tracked in real time
- Dual-boot OS support for Windows and Android
- Advanced Touch Handling
 - Moisture/Water Compensation
 - No false touch with condensation or water drop up to 22 mm diameter
 - One-finger tracking with condensation or water drop up to 22 mm diameter
 - Stylus Support
 - Supports passive stylus with 1 mm contact diameter, subject to configuration, stack up, and sensor design
 - Hover Support
 - Supports one-finger hover up to 25 mm detection and 20 mm tracking range
 - Glove Support
 - Supports multiple-finger glove touch up to 1.5 mm thickness
 - Supports single-touch gloved operation with various materials up to 5 mm thickness
- Touch Performance
 - Mutual capacitance and self capacitance measurements supported for touch detection
 - Response Times
 - Initial latency <10 ms for first touch from idle, subject to configuration
 - Atmel maXCharger® technology to combat ambient, charger noise, and power-line noise:
 - Up to 240 Vpp between 1 Hz and 1 kHz sinusoidal waveform
 - Up to 20 Vpp between 1 kHz and 1 MHz sinusoidal waveform
 - Scan Speed
 - Typical report rate for 10 touches ≥60 Hz
- Enhanced Algorithms
 - Lens bending algorithms to remove signal distortions
 - Touch suppression algorithms to remove unintentional touches
 - Palm Recovery Algorithm for quick restoration to normal state

- Panel / Cover Glass Support
 - Supports fully-laminated sensors, touch-on-lens stack-ups and on-cell designs
 - Works with PET or glass, including curved profiles
 - Glass from 0.55 mm to 2.5 mm, dependent on screen size and touch size
 - Plastic from 0.2 mm to 1.2 mm, dependent on screen size and touch size
 - Works with all proprietary sensor patterns recommended by Atmel
 - Compatible with True Single Layer designs
- Keys
 - Up to 32 nodes can be allocated as mutual capacitance sensor keys (subject to other configurations)
 - Adjacent Key Suppression® (AKS®) technology is supported for false touch prevention
- Power Saving
 - Programmable timeout for automatic transition from active to idle states
 - Pipelined analog sensing detection and digital processing to optimize system power efficiency
- Application Interfaces
 - I²C-compatible slave mode: Standard/Fast mode 400 kHz, Fast-plus mode 1 MHz, High-speed mode up to 3.4 MHz
 - HID-I²C interface for Microsoft® Windows® 8.x
 - Interrupt to indicate when a message is available
- Power Supply
 - Digital (V_{dd}) 3.3 V nominal
 - Analog (AV_{dd}) 3.3 V nominal
 - Host interface I/O voltage (V_{ddIO}) 1.8 V to 3.3 V nominal
 - High voltage internal X line drive (XV_{dd}) = 2 × V_{dd} (6.6 V), with internal voltage doubler
- Packages
 - 84-ball UFBGA 6 × 6 × 0.6 mm, 0.5 mm pitch
 - 84-ball XFLGA 6 × 6 × 0.5 mm, 0.5 mm pitch
- Environmental Conditions
 - Operating temperature –40°C to +85°C

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1. Overview of mXT641T

1.1 Introduction

The Atmel maXTouch family of touch controllers brings industry-leading capacitive touch performance to customer applications. The mXT641T features the latest generation of Atmel Adaptive Sensing technology that utilizes a hybrid mutual- and self-capacitive sensing system in order to deliver unparalleled touch features and a robust user experience.

- **Patented capacitive sensing method** – The mXT641T uses a unique charge-transfer acquisition engine to implement the Atmel-patented QMatrix[®] capacitive sensing method. Coupled with a state-of-the-art CPU, the entire touchscreen sensing solution can measure, classify and track number of individual finger touches with a high degree of accuracy in the shortest response time.
- **Capacitive Touch Engine (CTE)** – The mXT641T features an acquisition engine, which uses an optimal measurement approach to ensure almost complete immunity from parasitic capacitance on the receiver input lines. The engine includes sufficient dynamic range to cope with anticipated touchscreen self and mutual capacitances, which allows great flexibility for use with the Atmel proprietary sensor pattern designs. One- and two-layer ITO sensors are possible using glass or PET substrates.
- **Touch detection** – The mXT641T allows for both mutual- and self-capacitance measurements, with the self-capacitance measurements being used to augment the mutual-capacitance measurements to produce reliable touch information.

When self-capacitance measurements are enabled, touch classification is achieved using both mutual- and self-capacitance touch data. This has the advantage that both types of measurement systems can work together to detect touches under a wide variety of circumstances.

During idle mode, the device performs self-capacitance touch scans (interspersed with self-capacitance hover scans, if required). When a touch is detected, the device starts performing mutual-capacitance touch scans as well as self capacitance scans.

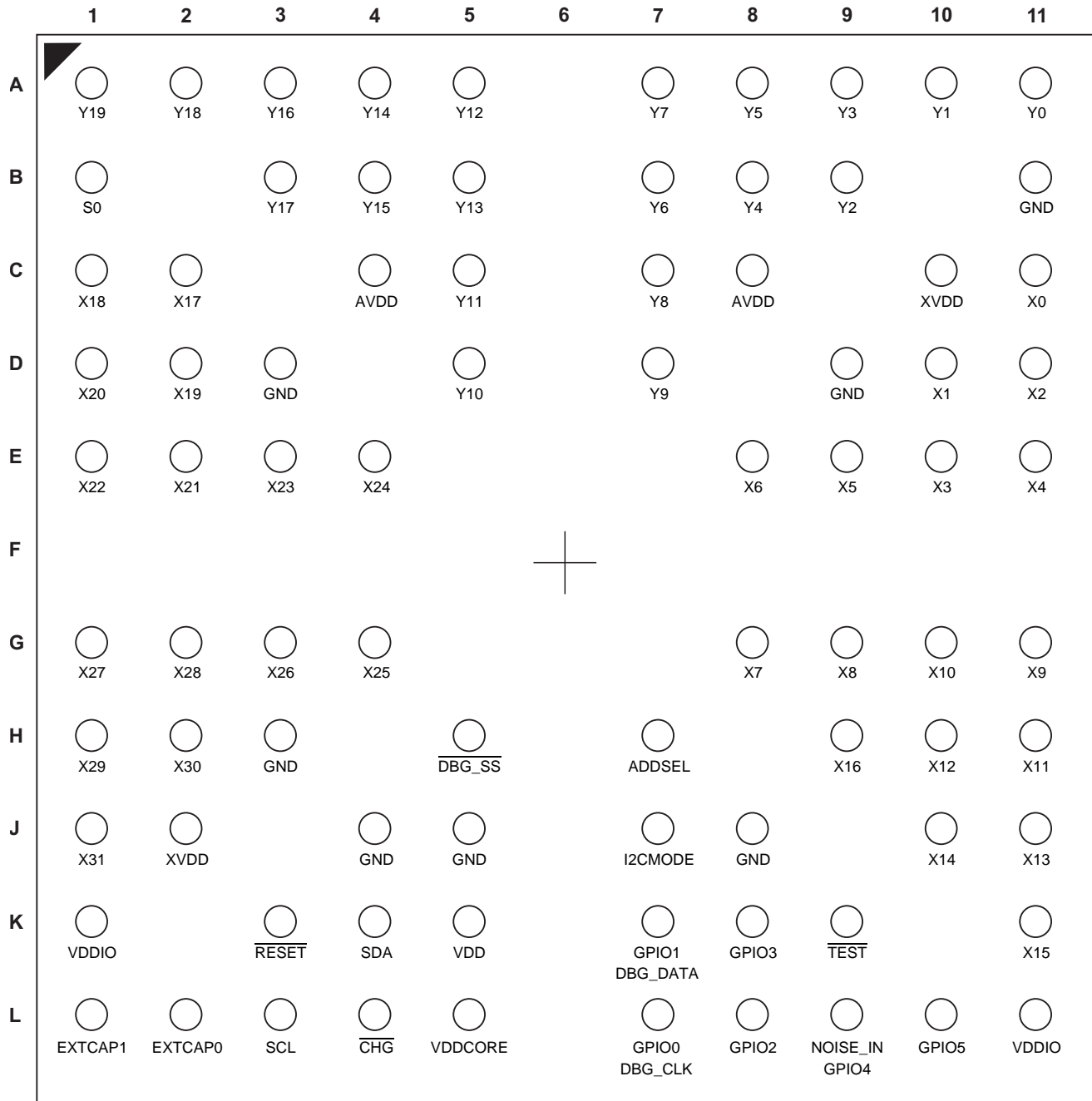
Mutual-capacitance touch data is used wherever possible to classify touches as this has greater granularity than self-capacitance measurements and provides positional information on touches. For this reason, multiple touches can only be determined by mutual-capacitance touch data. If the self-capacitance touch processing detects multiple touches, touchscreen processing is skipped until mutual-capacitance touch data is available.

Self-capacitance measurements, on the other hand, allow for the detection of single touches in extreme case, such as single thick-glove touches, when touches can only be detected by self-capacitance data and may be missed by mutual-capacitance touch detection.

- **Display Noise Cancellation** – A combination of analog circuitry, hardware noise processing, and firmware that combats display noise without requiring additional listening channels or synchronization to display timing. This enables the use of shieldless touch sensor stacks, including touch-on-lens.
- **Noise filtering** – Hardware noise processing in the capacitive touch engine provides enhanced autonomous filtering and allows a broad range of noise profiles to be handled. The result is good performance in the presence of charger and LCD noise.
- **Processing power** – The main CPU has two powerful microsequencer coprocessors under its control consuming low power. This system allows the signal acquisition, preprocessing, postprocessing and housekeeping to be partitioned in an efficient and flexible way.
- **Interpreting user intention** – The Atmel hybrid mutual- and self-capacitance method provides unambiguous multitouch performance. Algorithms in the mXT641T provide optimized touchscreen position filtering for the smooth tracking of touches, responding to a user's intended touches while preventing false touch triggered by ambient noise or conductive material on the sensor surface, such as water. The suppression of unintentional touches from the user's gripping fingers, resting palm or touching cheek or ear also help ensure that the user's intentions are correctly interpreted.

2. Connection and Configuration Information

2.1 Pin Configuration – UFBGA and XFLGA 84 Balls/Pads



Top View

Table 2-1. Pin Listing 84 UFBGA and XFLGA

Ball	Name	Type	Comments	If Unused, Connect To...
A1	Y19	S	Y line connection	Leave open
A2	Y18	S	Y line connection	Leave open
A3	Y16	S	Y line connection	Leave open
A4	Y14	S	Y line connection	Leave open
A5	Y12	S	Y line connection	Leave open
–				
A7	Y7	S	Y line connection	Leave open
A8	Y5	S	Y line connection	Leave open
A9	Y3	S	Y line connection	Leave open
A10	Y1	S	Y line connection	Leave open
A11	Y0	S	Y line connection	Leave open
B1	S0	S	Driven Shield signal; used as guard track between X/Y signals and ground	Leave open
–				
B3	Y17	S	Y line connection	Leave open
B4	Y15	S	Y line connection	Leave open
B5	Y13	S	Y line connection	Leave open
–				
B7	Y6	S	Y line connection	Leave open
B8	Y4	S	Y line connection	Leave open
B9	Y2	S	Y line connection	Leave open
–				
B11	GND	P	Ground	–
C1	X18	S	X line connection	Leave open
C2	X17	S	X line connection	Leave open
–				
C4	AVDD	P	Analog power	–
C5	Y11	S	Y line connection	Leave open
–				
C7	Y8	S	Y line connection	Leave open
C8	AVDD	P	Analog power	–
–				

Table 2-1. Pin Listing 84 UFBGA and XFLGA (Continued)

Ball	Name	Type	Comments	If Unused, Connect To...
C10	XVDD	P	X line drive power. For more information see Section 3.1.4 on page 14	–
C11	X0	S	X line connection	Leave open
D1	X20	S	X line connection	Leave open
D2	X19	S	X line connection	Leave open
D3	GND	P	Ground	–
–				
D5	Y10	S	Y line connection	Leave open
–				
D7	Y9	S	Y line connection	Leave open
–				
D9	GND	P	Ground	–
D10	X1	S	X line connection	Leave open
D11	X2	S	X line connection	Leave open
E1	X22	S	X line connection	Leave open
E2	X21	S	X line connection	Leave open
E3	X23	S	X line connection	Leave open
E4	X24	S	X line connection	Leave open
–				
E8	X6	S	X line connection	Leave open
E9	X5	S	X line connection	Leave open
E10	X3	S	X line connection	Leave open
E11	X4	S	X line connection	Leave open
–				
G1	X27	S	X line connection	Leave open
G2	X28	S	X line connection	Leave open
G3	X26	S	X line connection	Leave open
G4	X25	S	X line connection	Leave open
–				
G8	X7	S	X line connection	Leave open
G9	X8	S	X line connection	Leave open
G10	X10	S	X line connection	Leave open
G11	X9	S	X line connection	Leave open

Table 2-1. Pin Listing 84 UFBGA and XFLGA (Continued)

Ball	Name	Type	Comments	If Unused, Connect To...
H1	X29	S	X line connection	Leave open
H2	X30	S	X line connection	Leave open
H3	GND	P	Ground	–
–				
H5	$\overline{\text{DBG_SS}}$	I/O	Debug SS line	Pull up to VddIO
–				
H7	ADDSEL	I	I2C address select: Connect to GND for address 0x4A Pull up to VddIO for address 0x4B. For more information see Section 9.2 on page 29	–
–				
H9	X16	S	X line connection	Leave open
H10	X12	S	X line connection	Leave open
H11	X11	S	X line connection	Leave open
J1	X31	S	X line connection	Leave open
J2	XVDD	P	X line drive power. For more information see Section 3.1.4 on page 14	–
–				
J4	GND	P	Ground	–
J5	GND	P	Ground	–
–				
J7	I2CMODE	I	Selects I2C mode: HID-I2C: connect to GND I2C: Pull up to VddIO. For more information see Section 9.1 on page 29	–
J8	GND	P	Ground	–
–				
J10	X14	S	X line connection	Leave open
J11	X13	S	X line connection	Leave open
K1	VDDIO	P	Digital IO interface power	–
–				
K3	$\overline{\text{RESET}}$	I	Reset low. Connection to host system is recommended	Pull up to VddIO
K4	SDA	OD	Serial Interface Data	–
K5	VDD	P	Digital power	–

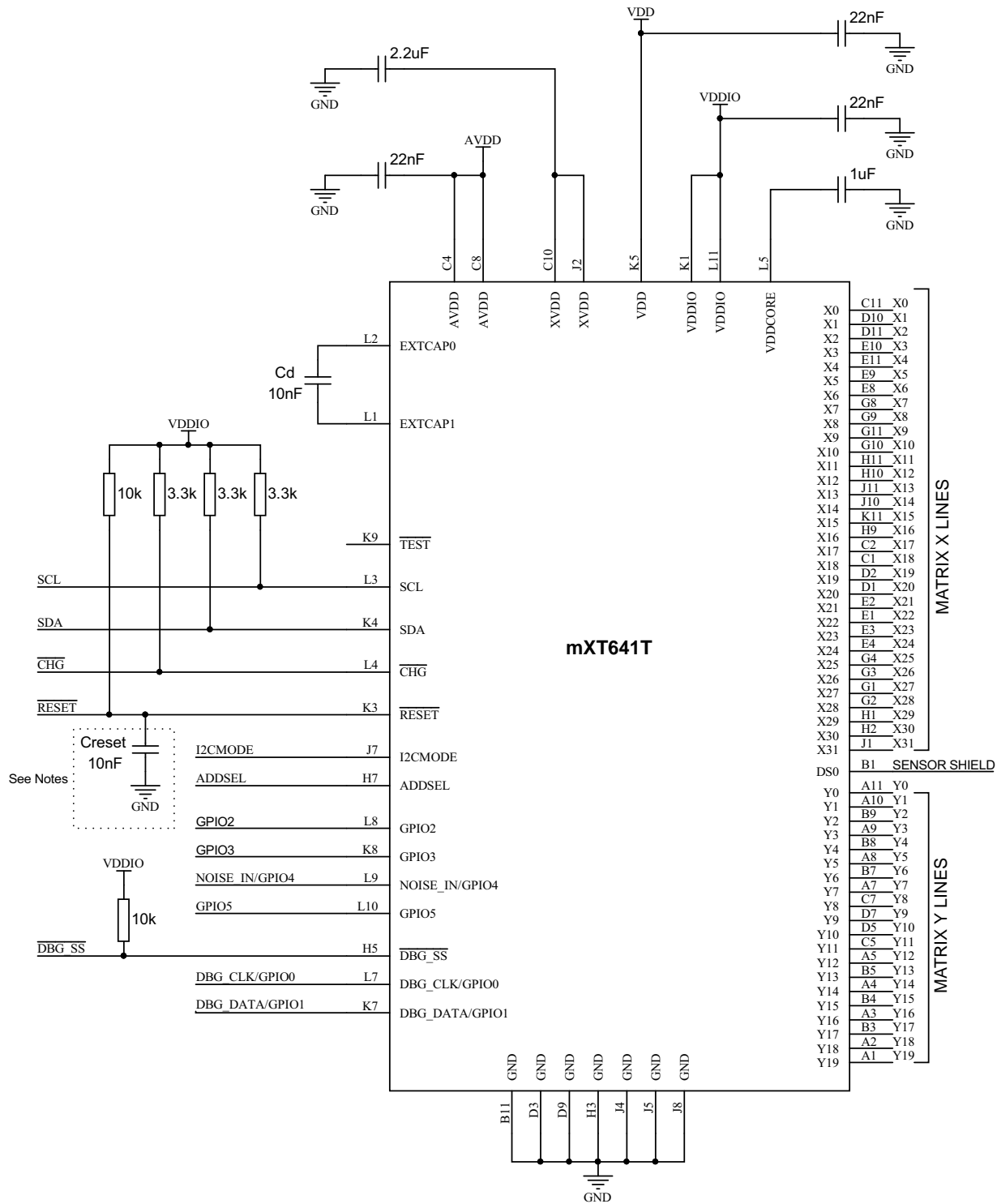
Table 2-1. Pin Listing 84 UFBGA and XFLGA (Continued)

Ball	Name	Type	Comments	If Unused, Connect To...
–				
K7	GPIO1 DBG_DATA	I/O	General purpose I/O Debug data output	Input: GND Output: leave open
K8	GPIO3	I/O	General purpose I/O	Input: GND Output: leave open
K9	$\overline{\text{TEST}}$	–	Reserved for factory use; leave open	Leave open
–				
K11	X15	S	X line connection	Leave open
L1	EXTCAP1	P	Connect to EXTCAP0 via capacitor; see schematic notes For more information see Section 3.1.4 on page 14	–
L2	EXTCAP0	P	Connect to EXTCAP1 via capacitor; see schematic notes For more information see Section 3.1.4 on page 14	–
L3	SCL	OD	Serial Interface Clock	–
L4	$\overline{\text{CHG}}$	OD	State change interrupt Note: Briefly set (~100 ms) as an input after power-up/reset for diagnostic purposes	Pull up to VddIO
L5	VDDCORE	P	Digital core power	–
–				
L7	GPIO0 DBG_CLK	I/O	General purpose I/O Debug clock input	Input: GND Output: leave open
L8	GPIO2	I/O	General purpose I/O	Input: GND Output: leave open
L9	NOISE_IN GPIO4	I/O	Noise present input General purpose I/O	Input: GND Output: leave open
L10	GPIO5	I/O	General purpose I/O	Leave open
L11	VDDIO	P	Digital IO interface power	–

Key:

I	Input only	O	Output only	I/O	Input or output
OD	Open drain output	P	Ground or power	S	Sense pin

3. Schematic



See Section 3.1 "Schematic Notes" on page 13

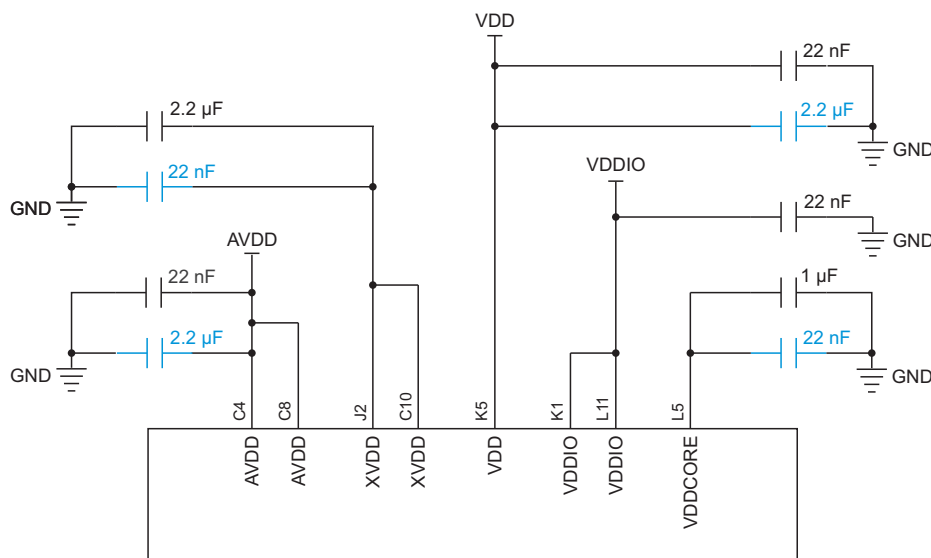
3.1 Schematic Notes

CAUTION: The device may be permanently damaged if any XVDD pin is shorted to ground or high current is drawn from it.

3.1.1 Decoupling capacitors:

1. Decoupling capacitors must be X7R or X5R and placed <5 mm away from the pins for which they act as bypass capacitors.
2. The schematics show the minimum capacitors required. If the ball configuration means that sharing a bypass capacitor is not possible, then the number of capacitors should be increased.
3. If the device is placed on the system board, the minimum number of capacitors required is as shown in the schematic on [page 12](#). Note that this requires that the voltage regulator supplies for AVDD/VDD and VDDIO are clean and noise free. It also assumes that the track length between the capacitors and on-board power supplies is < 50 mm.
4. If an active tail design is used, the voltage regulators are likely to be some distance from the device and it may be necessary to implement additional decoupling. In this case, a parallel combination of capacitors is recommended to give high and low frequency filtering, as shown in [Figure 3-1](#).

Figure 3-1. Additional Recommended Decoupling Capacitors



NOTE: Recommended additional decoupling capacitors are shown in blue

3.1.2 VDDCORE

VDDCore is internally generated from the Vdd 3.3 V power supply. To guarantee stability of the internal voltage regulator, a minimum value of 1 μF must be used for decoupling on VDDCORE.

3.1.3 RESET Line

The $\overline{\text{RESET}}$ line is shown on the schematics with a 10 nF capacitor (Creset) to ground. This capacitor is optional but may help if ESD issues are encountered.

3.1.4 Voltage Doubler

To use XVdd voltage doubler:

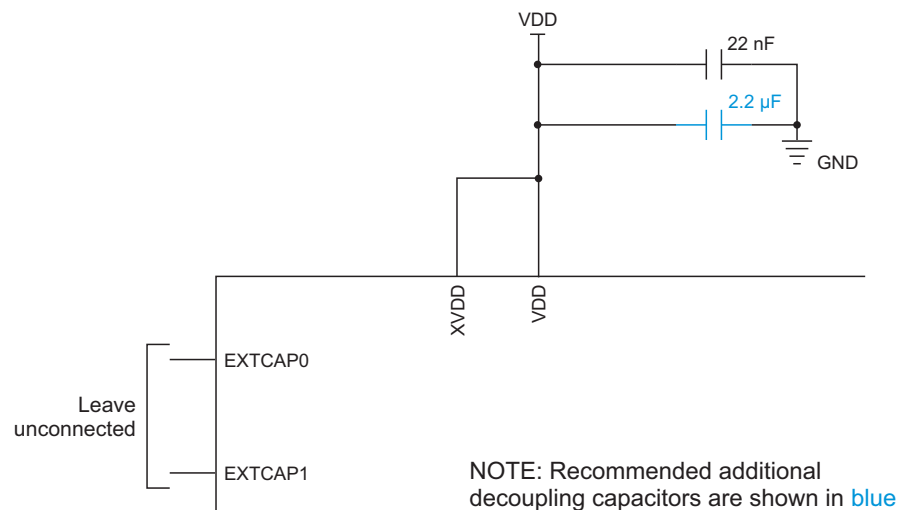
1. EXTCAP0 must be connected to EXTCAP1 via a capacitor (C_d) to provide XVdd voltage doubler mode.
2. The recommended value of the capacitor is 10 nF. Other values can be used if necessary after consultation with Atmel.

If XVdd voltage doubler is not required:

1. The capacitor can be omitted and EXTCAP0 and EXTCAP1 left unconnected.
2. XVDD line(s) should be connected to VDD.

These modifications are shown in [Figure 3-2](#).

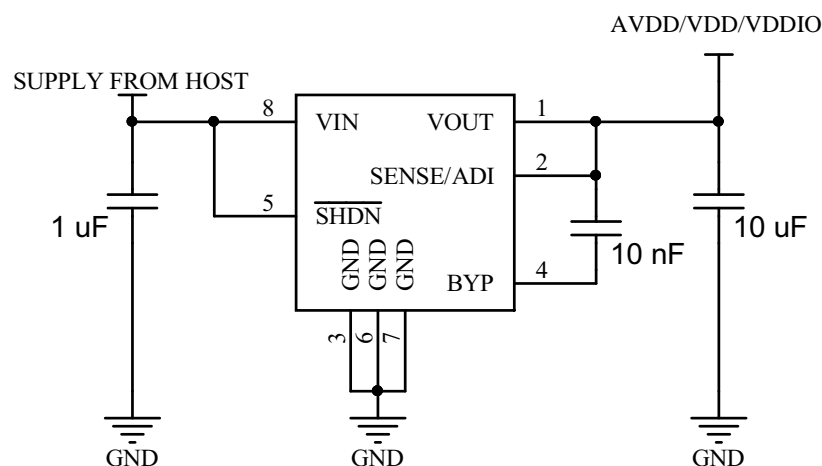
Figure 3-2. No voltage Doubler



3.1.5 Low Drop-Out Voltage Regulators (LDOs)

In applications where the VddIO supply is at the same voltage level as Vdd and AVdd (that is, 3.3 V) it is permissible to use a single LDO for all supply rails (AVDD/VDD/VDDIO). A suitable circuit is shown in [Figure 3-3](#).

Figure 3-3. Low Drop-Out Regulators



Where poor or inadequate tracking or decoupling leads to high noise levels on the supply rails, Atmel recommends that a separate low drop-out voltage regulator supply is used for the AVdd supply.

See [Section 4.4 on page 16](#) for further details. A list of approved regulators is given in [Table 4-1 on page 17](#).

4. Circuit Components

4.1 Decoupling Capacitors

Each power supply pin requires decoupling as described in [Section 3.1 on page 13](#). The capacitors should be ceramic X7R or X5R.

The PCB traces connecting the decoupling capacitors to the pins of the device must not exceed 10 mm in length. This limits any stray inductance that would reduce filtering effectiveness.

4.2 I²C Line Pull-up Resistors

The values for pull-up resistors on SDA and SCL need to be chosen to ensure rise times are within I²C specification – if the rise time is too long the overall clock rate will be reduced.

If using a VddIO at the low end of the allowable range it is likely that the pull-up resistor values will need to be reduced from those shown on the schematic.

4.3 Supply Quality

While the device has good Power Supply Rejection Ratio properties, poorly regulated and/or noisy power supplies can significantly reduce performance.

Always operate the device with a well-regulated and clean AVdd supply. It supplies the sensitive analog stages in the device.

4.4 Suggested Voltage Regulators

An LDO regulator should be chosen that provides adequate output capability, low noise, good load regulation and step response.

Suitable fixed output LDO devices are shown in [Table 4-1 on page 17](#).

With a single regulator, PCB layout is more critical than with multiple LDO regulators, and special care with the PCB layout should be taken. See [Section 12.5 on page 47](#) for information concerning PCB design with a single LDO.

4.4.1 Multiple Voltage Regulator Supply

The AVdd supply stability is critical for the device because this supply interacts directly with the analog front end. If noise problems exist when using a single LDO regulator, Atmel recommends that the supply for the analog section of the board be supplied by a regulator that is separate from the logic supply. This reduces the amount of noise injected into the sensitive, low signal level parts of the design.

4.4.2 Suggested Voltage Regulators

The voltage regulators listed in [Table 4-1](#) have been tested and found to work well with the mXT641T.

Table 4-1. Suitable LDO Regulators

Manufacturer	Device	Current Rating (mA)
Analog Devices	ADP122/ADP123	300
Diodes Inc.	AP2125	300
Diodes Inc.	AP7335	300
Linear Technology	LT1763CS8-3.3	500
NXP	LD6836	300
Texas Instruments	LP2981	100
Texas Instruments	LP3981	300
Texas Instruments	LP5996	150 / 300

1. Some manufacturers claim that minimal or no capacitance is required for correct regulator operation. However, in all cases, a minimum of a 1.0 μF ceramic, low ESR capacitor at the input and output of these devices should be used. The manufacturer's datasheets should always be referred to when selecting capacitors for these devices and the typical recommended values, types and dielectrics adhered to.

4.4.3 LDO Selection Criteria

The LDO devices in [Table 4-1](#) have been proved to provide satisfactory performance in Atmel maxTouch controllers, however, if it is desired to use an alternative LDO, certain performance criteria should be verified before using the device. These are:

- Stable with low value multi-layer ceramic capacitors on input and output – actual values will be device dependent, but it is good design practice to use values greater than the minimum specified in the LDO regulator data sheet
- Low output noise – less than 100 μV RMS over the range 10 Hz to 1 MHz
- Good load transient response – this should be less than 35 mV peak when a load step change of 100 mA is applied at the device output terminal
- Input supply requirement of between 4.5 V and 5.5 V
- Low quiescent current to improve battery life
- Thermal and current limit overload protection
- Ideally, select an LDO with common footprint, to allow interchanging between regulators

5. Touchscreen Basics

5.1 Sensor Construction

A touchscreen is usually constructed from a number of transparent electrodes. These are typically on a glass or plastic substrate. They can also be made using non-transparent electrodes, such as copper or carbon. Electrodes are constructed from Indium Tin Oxide (ITO) or metal mesh. Thicker electrodes yield lower levels of resistance (perhaps tens to hundreds of Ω /square) at the expense of reduced optical clarity. Lower levels of resistance are generally more compatible with capacitive sensing. Thinner electrodes lead to higher levels of resistance (perhaps hundreds to thousands of Ω /square) with some of the best optical characteristics.

Interconnecting tracks can cause problems. The excessive RC time constants formed between the resistance of the track and the capacitance of the electrode to ground can inhibit the capacitive sensing function. In such cases, the tracks should be replaced by screen printed conductive inks (non-transparent) outside the touchscreen viewing area.

5.2 Electrode Configuration

The specific electrode designs used in Atmel touchscreens are the subject of various patents and patent applications. Further information is available on request.

The device supports various configurations of electrodes as summarized in [Section 6. on page 20](#).

5.3 Scanning Sequence

All nodes are scanned in sequence by the device. There is a full parallelism in the scanning sequence to improve overall response time. The nodes are scanned by measuring capacitive changes at the intersections formed between the first X line and all the Y lines. Then the intersections between the next X line and all the Y lines are scanned, and so on, until all X and Y combinations have been measured.

The device can be configured in various ways. It is possible to disable some nodes so that they are not scanned at all. This can be used to improve overall scanning time.

5.4 Touchscreen Sensitivity

5.4.1 Adjustment

Sensitivity of touchscreens can vary across the extents of the electrode pattern due to natural differences in the parasitic capacitance of the interconnections, control chip, and so on. An important factor in the uniformity of sensitivity is the electrode design itself. It is a natural consequence of a touchscreen pattern that the edges form a discontinuity and hence tend to have a different sensitivity. The electrodes at the far edges do not have a neighboring electrode on one side and this affects the electric field distribution in that region.

A sensitivity adjustment is available for the whole touchscreen. This adjustment is a basic algorithmic threshold that defines when a node is considered to have enough signal change to qualify as being in detect.

5.4.2 Mechanical Stackup

The mechanical stackup refers to the arrangement of material layers that exist above and below a touchscreen. The arrangement of the touchscreen in relation to other parts of the mechanical stackup has an effect on the overall sensitivity of the screen. QMatrix technology has an excellent ability to operate in the presence of ground planes close to the sensor. QMatrix sensitivity is attributed more to the interaction of the electric fields between the transmitting (X) and receiving (Y) electrodes than to the surface area of these electrodes. For this reason, stray capacitance on the X or Y electrodes does not strongly reduce sensitivity.

Front panel dielectric material has a direct bearing on sensitivity. Plastic front panels are usually suitable up to about 1.2 mm, and glass up to about 2.5 mm (dependent upon the screen size and layout). The thicker the front panel, the lower the signal-to-noise ratio of the measured capacitive changes and hence the lower the resolution of the touchscreen. In general, glass front panels are near optimal because they conduct electric fields almost twice as easily as plastic panels.

Note: Care should be taken using ultra-thin glass panels as retransmission effects can occur, which can significantly degrade performance.

6. Sensor Layout

The specific electrode designs used in Atmel touchscreens are the subject of various patents and patent applications. Further information is available on request.

The physical matrix can be configured to have one or more touch objects. These are configured using the appropriate touch objects (Multiple Touch Touchscreen and Key Array). It is not mandatory to have all the allowable touch objects present. The objects are disabled by default so only those that you wish to use need to be enabled. Refer to the *mXT641T 1.0 Protocol Guide* for more information on configuring the touch objects.

The device supports various configurations of electrodes as summarized below:

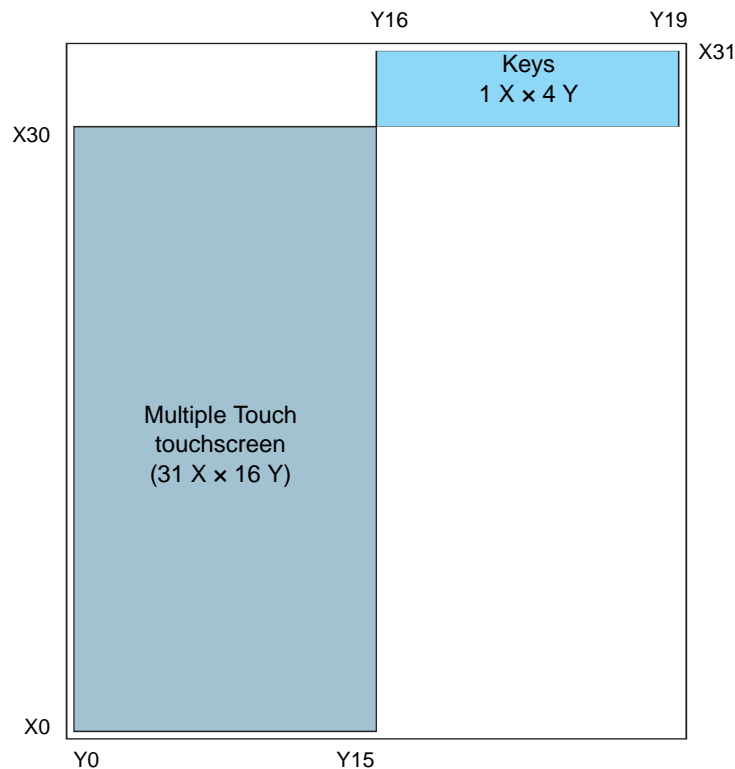
- Touchscreen: 32 X × 20 Y maximum (subject to other configurations)
- Keys: Up to 32 keys in an X/Y grid

When designing the physical layout of the touch panel, obey the following rules:

- Each touch object should be a regular rectangular shape in terms of the lines it uses.
- The touch objects cannot share X and Y lines
- It is recommended that the touchscreen should start at X0, Y0; if self-capacitance measurements are enabled, the touchscreen **must** start at X0, Y0
- It is recommended that the keys should occupy the highest X and Y lines

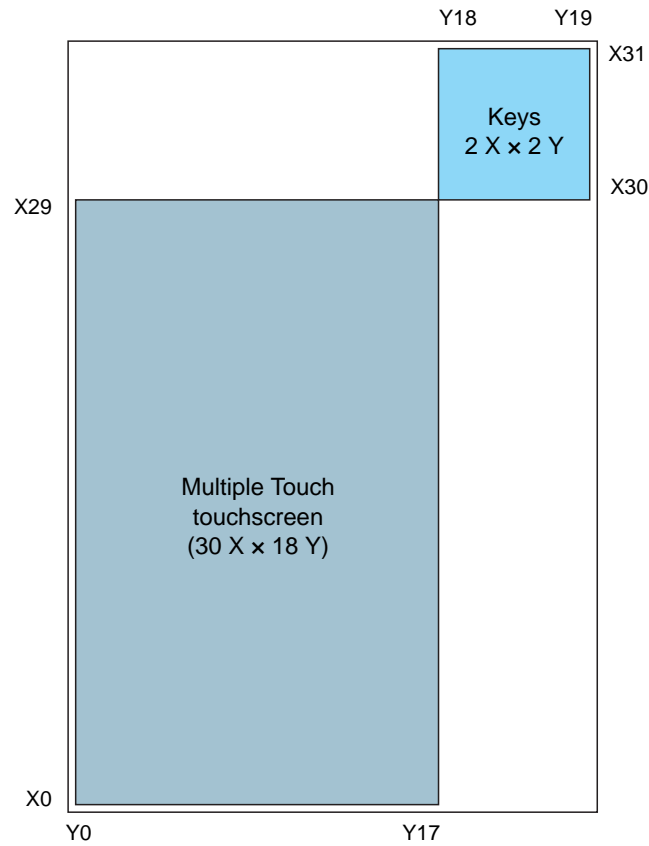
For optimal performance in terms of cycle time overhead, it is recommended that the number of X (drive) lines used is kept to the minimum and designs should favor using Y lines where possible. [Figure 6-1](#) shows an example layout for 1 X × 4 Y lines. Note that in this case using 1 X × 4 Y lines would give better performance than using 4 X × 1 Y lines.

Figure 6-1. Example Layout – Optimal Cycle Time



If, however, the intention is to preserve a larger touchscreen size and maintain an optimal aspect ratio, then using equal X and Y lines can be considered, as in [Figure 6-2](#).

Figure 6-2. Example Layout – Optimal Aspect Ratio

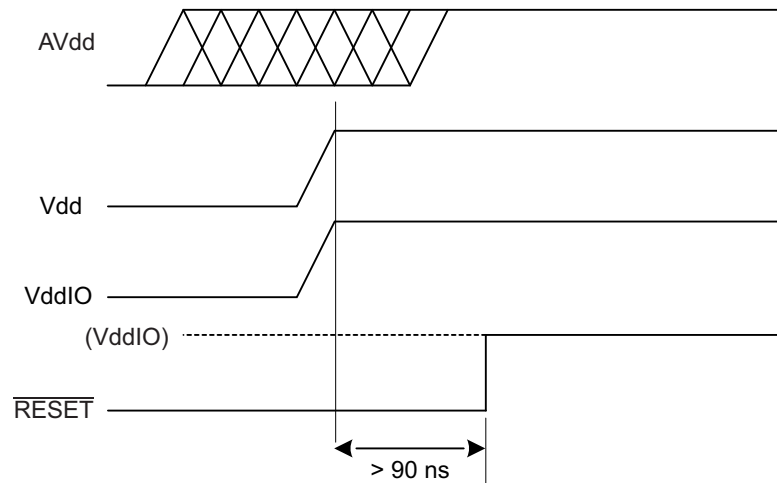


7. Power-up / Reset Requirements

There is an internal Power-on Reset (POR) in the device.

If an external reset is to be used the device must be held in $\overline{\text{RESET}}$ (active low) while the digital (Vdd) analog (AVdd) and I/O (VddIO) power supplies are powering up. The supplies must have reached their nominal values before the $\overline{\text{RESET}}$ signal is deasserted (that is, goes high). This is shown in [Figure 7-1](#). See [Section 15.2 on page 52](#) for nominal values for Vdd, VddIO, and AVdd.

Figure 7-1. Power Sequencing on the mXT641T

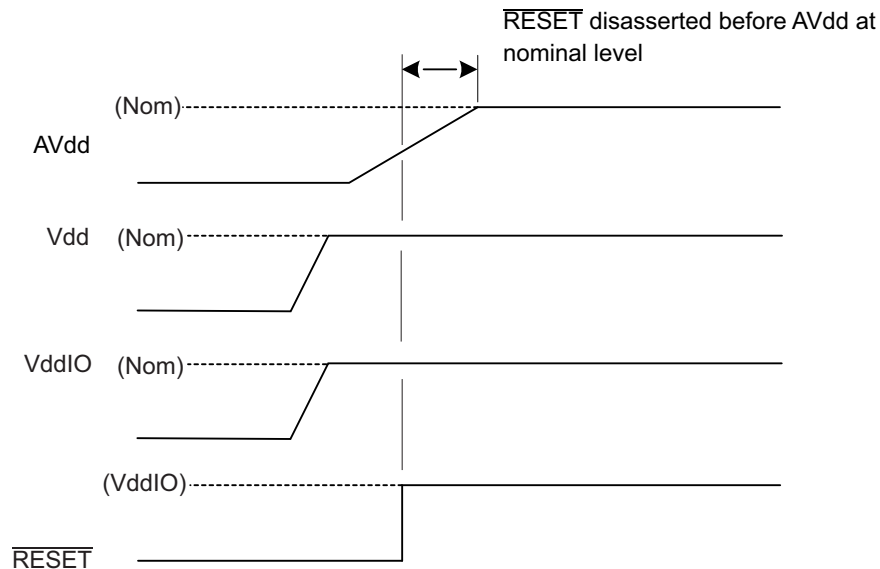


Note: When using external $\overline{\text{RESET}}$ at power-up, VddIO must not be enabled after Vdd

After power-up, the device takes 79 ms before it is ready to start communications.

If the $\overline{\text{RESET}}$ line is released before the AVdd supply has reached its nominal voltage (see [Figure 7-2 on page 23](#)), then some additional operations need to be carried out by the host. There are two options open to the host controller:

- Start the part in deep sleep mode and then send the command sequence to set the cycle time to wake the part and allow it to run normally. Note that in this case a calibration command is also needed.
- Send a reset command.

Figure 7-2. Power Sequencing on the mXT641T – Late rise on AVdd or XVdd

The $\overline{\text{RESET}}$ pin can be used to reset the device whenever necessary. The $\overline{\text{RESET}}$ pin must be asserted low for at least 90 ns to cause a reset. After releasing the $\overline{\text{RESET}}$ pin the device takes 79 ms before it is ready to start communications. It is recommended to connect the $\overline{\text{RESET}}$ pin to a host controller to allow it to initiate a full hardware reset without requiring a power-down.

Make sure that any lines connected to the device are below or equal to Vdd during power-up. For example, if $\overline{\text{RESET}}$ is supplied from a different power domain to the VDDIO pin, make sure that it is held low when Vdd is off. If this is not done, the $\overline{\text{RESET}}$ signal could parasitically couple power via the $\overline{\text{RESET}}$ pin into the Vdd supply.

Note that the voltage level on the RESET pin of the device must never exceed VddIO (digital supply voltage).

A software reset command can be used to reset the chip (refer to the Command Processor T6 object in the *mXT641T 1.0 Protocol Guide*). A software reset takes a maximum of 95 ms. After the chip has finished it asserts the $\overline{\text{CHG}}$ line to signal to the host that a message is available. The reset flag is set in the Message Processor object to indicate to the host that it has just completed a reset cycle. This bit can be used by the host to detect any unexpected brownout events. This allows the host to take any necessary corrective actions, such as reconfiguration.

A checksum check is performed on the configuration settings held in the nonvolatile memory. If the checksum does not match a stored copy of the last checksum, then this indicates that the settings have become corrupted. This is signaled to the host by setting the configuration error bit in the message data for the Command Processor T6 object (refer to the *mXT641T 1.0 Protocol Guide* for more information).

Note that the $\overline{\text{CHG}}$ line is briefly set as an input during power-up or reset. It is therefore particularly important that the line should be allowed to float high via the $\overline{\text{CHG}}$ line pull-up resistor during this period. It should not be driven by the host (see [Table 15.7.3 on page 58](#)).

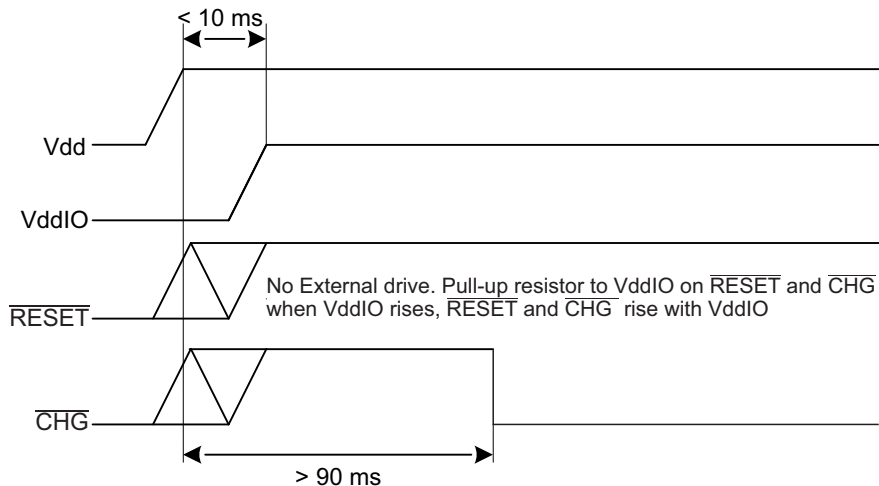
At power-on, the device performs a self-test routine to check for shorts that might cause damage to the device. Refer to the Self Test T25 object in the *mXT641T 1.0 Protocol Guide* for more details about this process.

7.1 Power-up and Reset Sequence – VddIO Enabled after Vdd

The Power-up sequence that can be used in applications where VddIO must be powered up after Vdd, is shown in [Figure 7-3](#).

In this case the communication interface to mXT is not driven by the host system. The $\overline{\text{RESET}}$ and $\overline{\text{CHG}}$ pins are connected to VddIO using suitable pull-up resistors. Vdd is powered up, followed by VddIO, no more than 10 ms after Vdd. Due to the pull-up resistors, $\overline{\text{RESET}}$ and $\overline{\text{CHG}}$ will rise with VddIO. The internal POR system ensures reliable boot up of the device and the $\overline{\text{CHG}}$ line will go low approximately 90 ms after Vdd to notify the host that the device is ready to start communication.

Figure 7-3. Power-up Sequence



7.1.1 Summary

The Power-up and RESET requirements for the maXTouch devices are summarised in the table below.

Condition	External $\overline{\text{RESET}}$	VddIO Delay (After Vdd)	AVdd Power-Up	Comments
1	Low at Power-up	0 ms	Before $\overline{\text{RESET}}$ is released	If AVdd bring-up is delayed then additional actions will be required by the host. See notes in Figure 7-1 on page 22
2	Not driven	<10 ms	Before VddIO	

8. Detailed Operation

8.1 Touch Detection

The mXT641T allows for both mutual and self capacitance measurements, with the self capacitance measurements being used to augment the mutual capacitance measurements to produce reliable touch information.

When self capacitance measurements are enabled, touch classification is achieved using both mutual and self capacitance touch data. This has the advantage that both types of measurement systems can work together to detect touches under a wide variety of circumstances.

Mutual capacitance touch data is used wherever possible to classify touches as this has greater granularity than self capacitance measurements and provides positional information on touches. Refer to the *mXT641T 1.0 Protocol Guide* for more information on measurements.

Self capacitance measurements, on the other hand, allow for the detection of single touches in extreme case, such as single thick glove touches, when touches can only be detected by self capacitance data and may be missed by mutual capacitance touch detection.

8.2 Operational Modes

The device operates in two modes: Active (touch detected) and Idle (no touches detected). Both modes operate as a series of burst cycles. Each cycle consists of a short burst (during which measurements are taken) followed by an inactive sleep period. The difference between these modes is the length of the cycles. Those in idle mode typically have longer sleep periods. The cycle length is configured using the IDLEACQINT and ACTVACQINT settings in the Power Configuration T7. In addition, an *Active to Idle Timeout* setting is provided.

Refer to the *mXT641T 1.0 Protocol Guide* for full information on how these modes operate, and how to use the settings provided.

8.3 Detection Integrator

The device features a touch detection integration mechanism. This acts to confirm a detection in a robust fashion. A counter is incremented each time a touch has exceeded its threshold and has remained above the threshold for the current acquisition. When this counter reaches a preset limit the sensor is finally declared to be touched. If, on any acquisition, the signal is not seen to exceed the threshold level, the counter is cleared and the process has to start from the beginning.

The detection integrator is configured using the appropriate touch objects (Multiple Touch Touchscreen T100, Key Array T15). Refer to the *mXT641T 1.0 Protocol Guide* for more information.

8.4 Sensor Acquisition

The maximum acquisition time for one X line on the mXT641T is 5 μ s. Care should be taken to ensure that the total time for one X line configured by the Acquisition Configuration T8 and CTE Configuration T46 objects do not exceed this (refer to the *mXT641T 1.0 Protocol Guide* for details on these objects).

8.5 Calibration

Calibration is the process by which a sensor chip assesses the background capacitance on each node. Nodes are only calibrated on reset and when:

- The node is enabled (that is, activated).

or

- The node is already enabled and one of the following applies:
 - The node is held in detect for longer than the Touch Automatic Calibration setting (refer to the *mXT641T 1.0 Protocol Guide* for more information on TCHAUTOCAL setting in the Acquisition Configuration object).

- The signal delta on a node is at least the touch threshold (TCHTHR) in the anti-touch direction, while it meets the criteria in the Touch Recovery Processes that results in a recalibration. (Refer to the *mXT641T 1.0 Protocol Guide* for objects Acquisition Configuration T8 and Self Capacitance Configuration T111).
- The host issues a recalibrate command.
- Certain configuration settings are changed.

A status message is generated on the start and completion of a calibration.

Note that the device performs a global calibration; that is, all the nodes are calibrated together.

8.6 Digital Filtering and Noise Suppression

The mXT641T supports on-chip filtering of the acquisition data received from the sensor. Specifically, the maXCharger T72 object provides an algorithm to suppress the effects of noise (for example, from a noisy charger plugged into the user's product). This algorithm can automatically adjust some of the acquisition parameters on-the-fly to filter the analog-to-digital conversions (ADCs) received from the sensor.

Additional noise suppression is provided by the Self Capacitance maXCharger T108 object. Similar in both design and configuration to the maXCharger T72 object, the Self Capacitance maXCharger T108 object is the noise suppression interface for self capacitance touch measurements.

Noise suppression is triggered when a noise source is detected.

- A hardware trigger can be implemented using the NOISE_IN pin.
- The host driver code can indicate when a noise source is present.
- The noise suppression is also triggered based on the noise levels detected using internal line measurements. The maXCharger T72 and Self Capacitance maXCharger T108 object selects the appropriate controls to suppress the noise present in the system.

Refer to the *mXT641T 1.0 Protocol Guide* for more information on the maXCharger T72 and Self Capacitance maXCharger T108 objects.

8.7 Shieldless Support and Display Noise Suppression

The mXT641T can support shieldless sensor design even with a noisy LCD by using the following features.

- **Optimal Integration:** This feature is not filtering as such, but enables the user to use a shorter integration window. The integration window optimizes the amount of charge collected against the amount of noise collected, to ensure an optimal SNR. This feature also benefits the system in the presence of an external noise source. This feature is configured using the Shieldless T56 object. Refer to the *mXT641T 1.0 Protocol Guide* for more information
- **Display noise suppression:** This feature is based on filtering provided by the Lens Bending T65 object (See [Section 8.10 on page 27](#)). This feature allows the device to overcome display noise simultaneously with external noise. Refer to the *mXT641T 1.0 Protocol Guide* for more information

8.8 Retransmission Compensation

The device can limit the undesirable effects on the mutual capacitance touch signals caused by poor device coupling to ground, such as poor sensitivity and touch break-up. This is achieved using the Retransmission Compensation T80 object. This object can be configured to allow the touchscreen to compensate for signal degradation due to these undesirable effects. If self capacitance measurements are also scheduled, the Retransmission Compensation T80 object will use the resultant data to enhance the compensation process.

The Retransmission Compensation T80 object is also capable of compensating for water presence on the sensor if self capacitance measurements are scheduled. In this case, both mutual capacitance and self capacitance measurements are used to detect moisture and then, once moisture is detected, self capacitance measurements are used to detect single touches in the presence of moisture.

Refer to the *mXT641T 1.0 Protocol Guide* for more information on the Retransmission Compensation T80 object.

8.9 Grip Suppression

The device has two grip suppression mechanisms to suppress false detections from a user's grip.

Mutual grip suppression works by specifying a boundary around a touchscreen, within which touches can be suppressed whilst still allowing touches in the center of the touchscreen. This ensures that a "rolling" hand touch (such as when a user grips a mobile device) is suppressed. A "real" (finger) touch towards the center of the screen is allowed.

Mutual grip suppression is configured using the Grip Suppression T40 object. There is one instance of the Grip Suppression T40 object for each Multiple Touch Touchscreen T100 object present on the device.

Refer to the *mXT641T 1.0 Protocol Guide* for more information on the Grip Suppression T40 and the Self Capacitance Grip Suppression T112 objects.

Self Capacitance grip suppression works by looking for characteristic shapes in the self capacitance measurement along the touchscreen boundary, and thereby distinguishing between a grip and a touch further into the sensor.

8.10 Lens Bending

The device supports algorithms to eliminate disturbances from the measured signal and also to measure the bend component.

When the sensor suffers from the screen deformation (lens bending) the signal values acquired by normal procedure are corrupted by the disturbance component (bend). The amount of bend depends on:

- The mechanical and electrical characteristics of the sensor
- The amount and location of the force applied by the user touch to the sensor

The Lens Bending T65 object measures the bend component and compensates for any distortion caused by the bend. As the bend component is primarily influenced by the user touch force, it can be used as a secondary source to identify the presence of a touch. The additional benefit of the Lens Bending T65 object is that it will eliminate LCD noise as well. Refer to the *mXT641T 1.0 Protocol Guide* for more information on the Lens Bending T65 object.

8.11 Glove Detection

The device has glove detection algorithms that process the measurement data received from the touchscreen classifying touches as potential gloved touches.

The Glove Detection T78 object is used to detect glove touches. In Normal Mode the Glove Detection T78 object applies vigorous glove classification to small signal touches to minimize the effect of unintentional hovering finger reporting.

Once a gloved touch is found, the Glove Detection T78 object enters Glove Confidence Mode. In this mode the device expects the user to be wearing gloves so the classification process is much less stringent.

Refer to the *mXT641T 1.0 Protocol Guide* for more information on the Glove Detection T78 object.

8.12 Stylus Support

The mXT641T allows for the particular characteristics of passive stylus touches, whilst still allowing conventional finger touches to be detected. The touch sensitivity and threshold controls for stylus touches are configured separately from those for conventional finger touches so that both types of touches can be accommodated.

Stylus support ensures that the small touch area of a stylus registers as a touch, as this would otherwise be considered too small for the touchscreen. Additionally, there are controls to distinguish a stylus touch from an unwanted approaching finger (such as on the hand holding the stylus).

Passive stylus touches are configured by the Stylus T47 object. There is one instance of the Stylus T47 object for each Multiple Touch Touchscreen T100 object present on the device.

Refer to the *mXT641T 1.0 Protocol Guide* for more information on configuring a stylus.

8.13 Hover Support

The mXT641T supports hover and is configured using the Touchscreen Hover Configuration T101 and the Self Capacitance Configuration T111 objects. The mXT641T allow for the configuration of both the hover measurements and also the data that defines the hover touch detection and post processing. The Hover touch status information can be accessed over the I²C interrupt-driven mechanism (Object Based Protocol) interface via the touch messages generated by the Multiple Touch Touchscreen T100 object, which must be disabled if the digitizer HID is used to access touch status information. Refer to the *mXT641T 1.0 Protocol Guide* for more information on the Touchscreen Hover Configuration T101 and Self Capacitance Configuration T111 objects.

8.14 Unintentional Touch Suppression

The Touch Suppression T42 object provides a mechanism to suppress false detections from unintentional touches from a large body area, such as from a face, ear or palm. The Touch Suppression T42 object also provides Maximum Touch Suppression to suppress all touches if more than a specified number of touches has been detected. There is one instance of the Touch Suppression T42 object for each Multiple Touch Touchscreen T100 object present on the device. Refer to the *mXT641T 1.0 Protocol Guide* for more information on the Touch Suppression T42 object.

8.15 Adjacent Key Suppression Technology

Adjacent Key Suppression (AKS) technology is a patented method used to detect which touch object is touched when objects are located close together. A touch in a group of AKS objects is only indicated on the object in that group that is touched first. This is assumed to be the intended object. Once an object in an AKS group is in detect, there can be no further detections within that group until the object is released. Objects can be in more than one AKS group.

Note that AKS technology works best when it operates in conjunction with a detect integration setting of several acquisition cycles.

The device has two levels of AKS. The first level works between the touch objects (Multiple Touch Touchscreen T100 and Key Array T15). The touch objects are assigned to AKS groups. If a touch occurs within one of the touch objects in a group, then touches within other objects inside that group are suppressed. For example, if a touchscreen and a Key Array are placed in the same AKS group, then a touch in the touchscreen will suppress touches in the Key Array, and vice versa.

The second level of AKS is internal AKS within an individual Key Array object (note that internal AKS is not present on other types of touch objects, only a Key Array T15). If internal AKS is enabled, then when one key is touched, touches on all the other keys within the Key Array are suppressed.

AKS is configured using the touch objects (Multiple Touch Touchscreen T100 or Key Array T15).

Refer to the *mXT641T 1.0 Protocol Guide* for more information.

Note: If a touch is in detect and then AKS is enabled, that touch will not be forced out of detect. It will not go out of detect until the touch is released. AKS will then operate normally. This applies to both levels of AKS.

8.16 GPIO Pins

The mXT641T has 6 GPIO pins. The pins can be set to be either an input or an output, as required. Note that unused GPIO pins can be left externally unconnected as long as they are given a defined state by using the GPIO/PWM Configuration T19 object. With the GPIO/PWM Configuration T19 object, an unused GPIO pin can be either set to Input mode, with internal pull-up, or Output mode.

By default GPIO pins are set to be inputs. If not used they should be connected to GND. Alternatively, they can be set as outputs using the GPIO/PWM Configuration T19 object and left open.

9. Host Communications

9.1 I²C Mode Selection (I2CMODE Pin)

The selection of the I²C or the HID-I²C mode is determined by the I2CMODE pin:

- If the I2CMODE pin is low (GND), HID-I²C mode is selected.
- If the I2CMODE pin is high (VddIO), I²C mode is selected.

9.2 I²C Address Selection (ADDSEL Pin)

The I²C address is selected by connecting the ADDSEL pin according to [Table 9-1](#).

Table 9-1. I²C Address Selection

ADDSEL	I ² C Address
Connected to GND	0x4A
Pulled up to VddIO	0x4B

10. I²C Communications

The device can use an I²C interface for communication. See [Appendix B. on page 67](#) for details of the I²C protocol.

The I²C interface is used in conjunction with the $\overline{\text{CHG}}$ line. The $\overline{\text{CHG}}$ line going active signifies that a new data packet is available. This provides an interrupt-style interface and allows the device to present data packets when internal changes have occurred.

To use the device in I²C mode, the I2CMODE pin should be pulled high.

10.1 I²C Addresses

The device supports two I²C device addresses that are selected using the ADDSEL line at start up. The two internal I²C device addresses are 0x4A and 0x4B. The selection of the address (and the communication mode) is described in [Section 9.2 on page 29](#). These are shifted left to form the SLA+W or SLA+R address when transmitted over the I²C interface, as shown in [Figure 10-1](#).

Table 10-1. Format of an I²C Address

Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
Address: 0x4A or 0x4B							Read/write

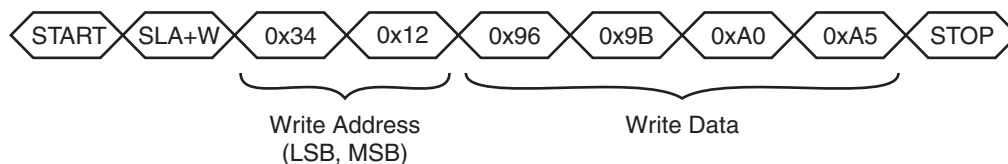
10.2 Writing To the Device

A WRITE cycle to the device consists of a START condition followed by the I²C address of the device (SLA+W). The next two bytes are the address of the location into which the writing starts. The first byte is the Least Significant Byte (LSByte) of the address, and the second byte is the Most Significant Byte (MSByte). This address is then stored as the address pointer.

Subsequent bytes in a multi-byte transfer form the actual data. These are written to the location of the address pointer, location of the address pointer + 1, location of the address pointer + 2, and so on. The address pointer returns to its starting value when the WRITE cycle STOP condition is detected.

[Figure 10-1](#) shows an example of writing four bytes of data to contiguous addresses starting at 0x1234.

Figure 10-1. Example of a Four-byte Write Starting at Address 0x1234

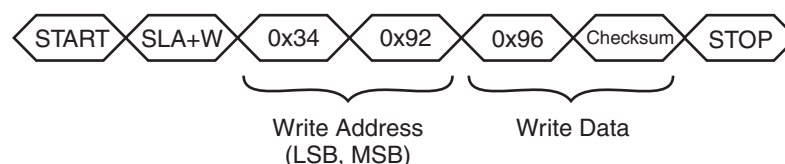


10.3 I²C Writes in Checksum Mode

In I²C checksum mode an 8-bit CRC is added to all I²C writes. The CRC is sent at the end of the data write as the last byte before the STOP condition. All the bytes sent are included in the CRC, including the two address bytes. Any command or data sent to the device is processed even if the CRC fails.

To indicate that a checksum is to be sent in the write, the most significant bit of the MSByte of the address is set to 1. For example, the I²C command shown in [Figure 10-2](#) writes a value of 150 (0x96) to address 0x1234 with a checksum. The address is changed to 0x9234 to indicate checksum mode.

Figure 10-2. Example of a Write To Address 0x1234 With a Checksum



10.4 Reading From the Device

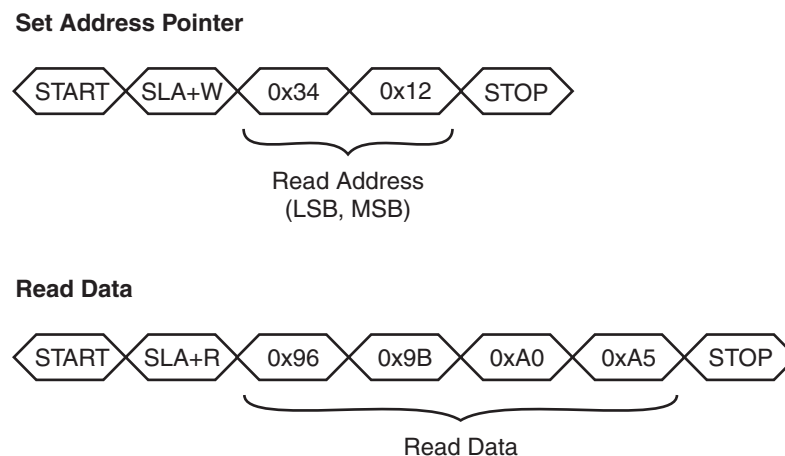
Two I²C bus activities must take place to read from the device. The first activity is an I²C write to set the address pointer (LSByte then MSByte). The second activity is the actual I²C read to receive the data. The address pointer returns to its starting value when the read cycle NACK is detected.

It is not necessary to set the address pointer before every read. The address pointer is updated automatically after every read operation. The address pointer will be correct if the reads occur in order. In particular, when reading multiple messages from the Message Processor T5 object, the address pointer is automatically reset to allow continuous reads (see Section 10.5).

The WRITE and READ cycles consist of a START condition followed by the I²C address of the device (SLA+W or SLA+R respectively).

Figure 10-3 shows the I²C commands to read four bytes starting at address 0x1234.

Figure 10-3. Example of a Four-byte Read Starting at Address 0x1234



10.5 Reading Status Messages with DMA

The device facilitates the easy reading of multiple messages using a single continuous read operation. This allows the host hardware to use a direct memory access (DMA) controller for the fast reading of messages, as follows:

1. The host uses a write operation to set the address pointer to the start of the Message Count T44 object, if necessary ⁽¹⁾. If a checksum is required on each message, the most significant bit of the MSByte of the read address must be set to 1.
2. The host starts the read operation of the message by sending a START condition.
3. The host reads the Message Count T44 object (one byte) to retrieve a count of the pending messages (refer to the *mXT641T 1.0 Protocol Guide* for details).
4. The host calculates the number of bytes to read by multiplying the message count by the size of the Message Processor T5 object ⁽²⁾.
5. Note that the size of the Message Processor T5 object as recorded in the Object Table includes a checksum byte. If a checksum has not been requested, one byte should be deducted from the size of the object. That is: number of bytes = count × (size – 1).
6. The host reads the calculated number of message bytes. It is important that the host does *not* send a STOP condition during the message reads, as this will terminate the continuous read operation and reset the address pointer. No START and STOP conditions must be sent between the messages.

1. The STOP condition at the end of the read resets the address pointer to its initial location, so it may already be pointing at the Message Count T44 object following a previous message read.

2. The host should have already read the size of the Message Processor T5 object in its initialization code.

- The host sends a STOP condition at the end of the read operation after the last message has been read. The NACK condition immediately before the STOP condition resets the address pointer to the start of Message Count T44 object.

Figure 10-4 shows an example of using a continuous read operation to read three messages from the device without a checksum. Figure 10-5 on page 33 shows the same example with a checksum.

Figure 10-4. Continuous Message Read Example – No Checksum

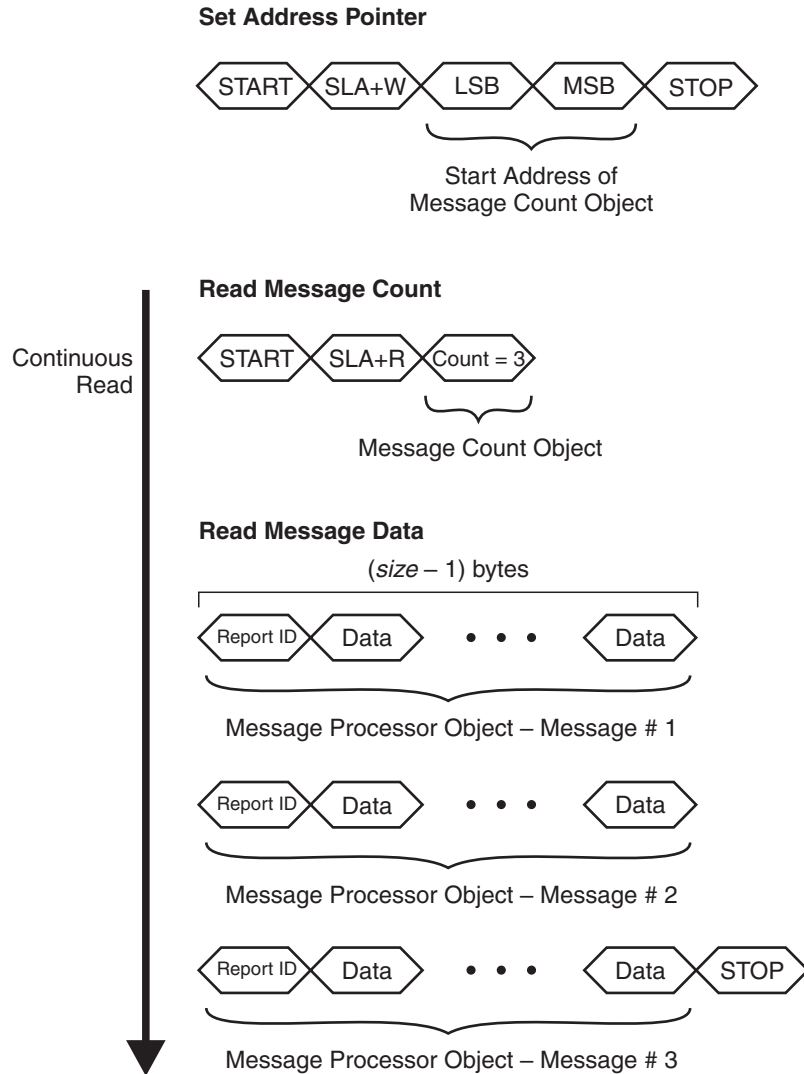
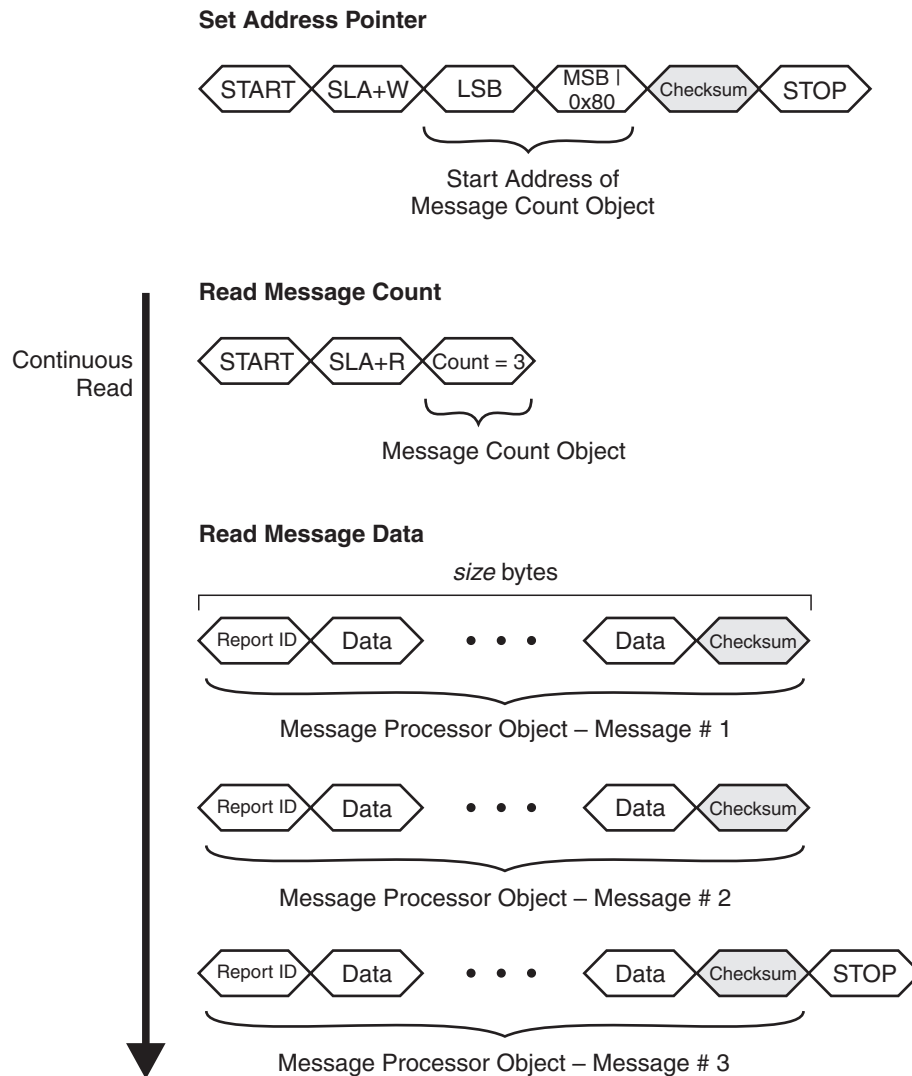


Figure 10-5. Continuous Message Read Example – I²C Checksum Mode

There are no checksums added on any other I²C reads. An 8-bit CRC can be added, however, to all I²C writes, as described in [Section 10.3 on page 30](#).

An alternative method of reading messages using the $\overline{\text{CHG}}$ line is given in [Section 10.6](#).

10.6 $\overline{\text{CHG}}$ Line

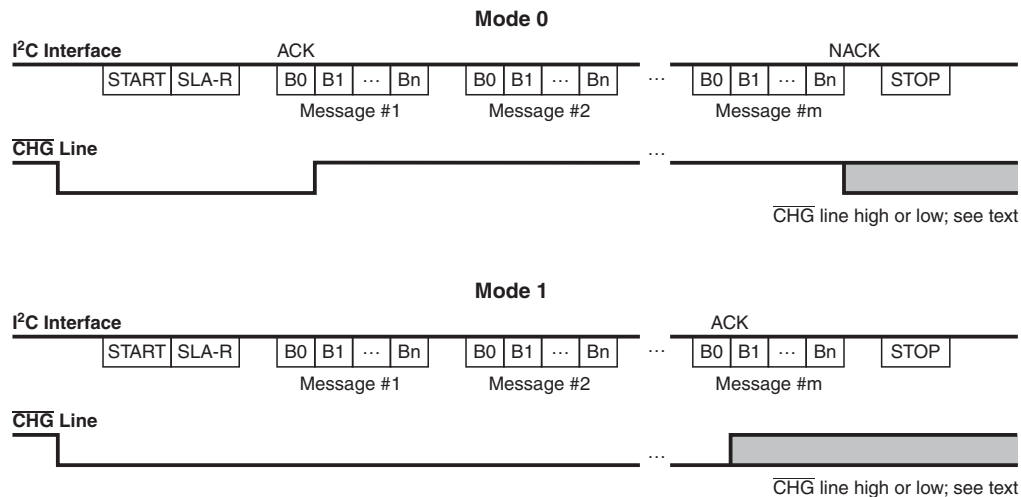
The $\overline{\text{CHG}}$ line is an active-low, open-drain output that is used to alert the host that a new message is available in the Message Processor T5 object. This provides the host with an interrupt-style interface with the potential for fast response times. It reduces the need for wasteful I²C communications.

The $\overline{\text{CHG}}$ line remains low as long as there are messages to be read. The host should be configured so that the $\overline{\text{CHG}}$ line is connected to an interrupt line that is level-triggered. The host should not use an edge-triggered interrupt as this means adding extra software precautions.

The $\overline{\text{CHG}}$ line should be allowed to float during normal usage. This is particularly important after power-up or reset (see [Section 7. on page 22](#)).

A pull-up resistor is required, typically 3.3 k Ω to VddIO.

The $\overline{\text{CHG}}$ line operates in two modes, as defined by the Communications Configuration T18 object (refer to the *mXT641T 1.0 Protocol Guide*).

Figure 10-6. $\overline{\text{CHG}}$ Line Modes for I²C-compatible Transfers**In Mode 0:**

1. The $\overline{\text{CHG}}$ line goes low to indicate that a message is present.
2. The $\overline{\text{CHG}}$ line goes high when the first byte of the first message (that is, its report ID) has been sent and acknowledged (ACK sent) and the next byte has been prepared in the buffer.
3. The STOP condition at the end of an I²C transfer causes the $\overline{\text{CHG}}$ line to stay high if there are no more messages. Otherwise the $\overline{\text{CHG}}$ line goes low to indicate a further message.

Mode 0 allows the host to continually read messages. Messaging reading ends when a report ID of 255 (“invalid message”) is received. Alternatively the host ends the transfer by sending a NACK after receiving the last byte of a message, followed by a STOP condition. If and when there is another message present, the $\overline{\text{CHG}}$ line goes low, as in step 1. In this mode the state of the $\overline{\text{CHG}}$ line does not need to be checked during the I²C read.

In Mode 1:

1. The $\overline{\text{CHG}}$ line goes low to indicate that a message is present.
2. The $\overline{\text{CHG}}$ line remains low while there are further messages to be sent after the current message.
3. The $\overline{\text{CHG}}$ line goes high again only once the first byte of the last message (that is, its report ID) has been sent and acknowledged (ACK sent) and the next byte has been prepared in the output buffer.

Mode 1 allows the host to continually read the messages until the $\overline{\text{CHG}}$ line goes high, and the state of the $\overline{\text{CHG}}$ line determines whether or not the host should continue receiving messages from the device.

Note: The state of the $\overline{\text{CHG}}$ line should be checked only between messages and not between the bytes of a message. The precise point at which the $\overline{\text{CHG}}$ line changes state cannot be predicted and so the state of the $\overline{\text{CHG}}$ line cannot be guaranteed between bytes.

The Communications Configuration T18 object can be used to configure the behavior of the $\overline{\text{CHG}}$ line. In addition to the $\overline{\text{CHG}}$ line operation modes described above, this object allows the use of edge-based interrupts, as well as direct control over the state of the $\overline{\text{CHG}}$ line. Refer to the *mXT641T 1.0 Protocol Guide* for more information.

10.7 SDA, SCL

The I²C bus transmits data and clock with SDA and SCL, respectively. These are open-drain. The device can only drive these lines low or leave them open. The termination resistors (Rp) pull the line up to Vdd if no I²C device is pulling it down.

The termination resistors should be chosen so that the rise times on SDA and SCL meet the I²C specifications for the interface speed being used, bearing in mind other loads on the bus (see [Section 15.9 on page 59](#)).

10.8 Clock Stretching

The device supports clock stretching in accordance with the I²C specification. It may also instigate a clock stretch if a communications event happens during a period when the device is busy internally. The maximum clock stretch is approximately 10 – 15 ms.

The device has an internal bus monitor that can reset the internal I²C hardware if SDA or SCL is stuck low for more than 200 ms. This means that if a prolonged clock stretch of more than 200 ms is seen by the device, then any ongoing transfers with the device may be corrupted. The bus monitor is enabled or disabled using the Communications Configuration T18 object. Refer to the *mXT641T 1.0 Protocol Guide* for more information.

11. HID-I²C Communications

The device is an HID-I²C device presenting two Top-level Collections (TLCs):

- **Generic HID-I²C** – Provides a generic HID-I²C interface that allows the host to communicate with the device using the object-based protocol (OBP).
- **Digitizer HID-I²C** – Supplies touch information to the host. This interface is supported by Microsoft Windows 8 and 8.1 without the need for additional software.

See [Section 9. on page 29](#) for information on selecting HID-I²C mode.

Other features are identical to standard I²C communication described in [Section 10. on page 30](#).

Refer to the Microsoft HID-I²C documentation, *HID Over I²C Protocol Specification – Device Side*, for information on the HID-I²C specification.

11.1 I²C Addresses

See [Section 10.1 on page 30](#).

11.2 Device

The device is compliant with HID-I²C specification V1.0. It has the following specification:

Vendor ID: 0x03EB (Atmel)
 Product ID: 0x214D (mXT641T)
 Version: 16-bit Version & Build Identifier in the form 0xVVBB, where:
 VV = Version Major (Upper 4 bits) / Minor (Lower 4 bits)
 BB = Build number in BCD format

HID descriptor address: 0x0000

11.3 HID Descriptor

The host should read the HID descriptor on initialization to ascertain the key attribute of the HID device. These include the report description and the report ID to be used for communication with the HID device. The HID descriptor address is 0x0000.

Note that the host driver must not make any assumptions about the report packet formats, data locations or report IDs. These must be read from the HID descriptor as they may change in future versions of the firmware.

For more information on how to read the HID descriptor, refer to the Microsoft HID-I²C documentation.

11.4 HID-I²C Report IDs

[Table 11-1](#) describes the HID-I²C report IDs used in reports sent to the host.

Table 11-1. HID-I²C Report IDs

Report ID	Description	Top-level Collection
0x06	Object Protocol (OBP) command and response (see Section 11.5 on page 37)	Generic HID-I ² C
0x01	Touch report (see Section 11.6.1 on page 41)	Digitizer HID-I ² C
0x02	Maximum Touches (Surface Contacts) report (see Section 11.6.3 on page 44)	Digitizer HID-I ² C
0x05	Touch Hardware Quality Assurance (THQA) report (see Section 11.6.4 on page 44)	Digitizer HID-I ² C

11.5 Generic HID-I²C

The Generic HID-I²C TLC supports an input report for receiving data from the device and an output report for sending data to the device.

Commands are sent by the host using the output reports. Responses from the device are sent using input reports.

Supported commands are:

- Read/Write Memory Map
- Send Auto-return Messages

The HID-I²C Report ID used is that for Object Protocol commands and responses; see [Table 11-1 on page 36](#) for the value.

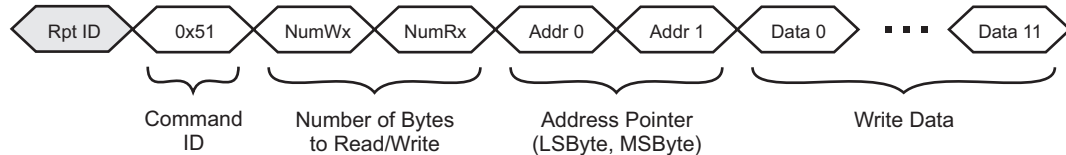
11.5.1 Read/Write Memory Map

11.5.1.1 Introduction

This command is used to carry out a write/read operation on the memory map of the device.

The command packet has the generic format given in [Figure 11-1](#). The following sections give examples on using the command to write to the memory map and to read from the memory map.

Figure 11-1. Generic Command Packet Format

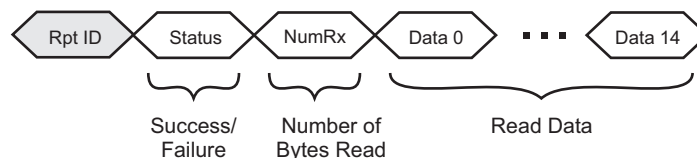


In [Figure 11-1](#):

- **Rpt ID** is the HID-I²C Report ID used for Object Protocol commands and responses (see [Table 11-1 on page 36](#)).
- **NumWx** is the number of data bytes to write to the memory map (may be zero). If the address pointer is being sent, this must include the size of the address pointer.
- **NumRx** is the number of data bytes to read from the memory map (may be zero).
- **Addr 0** and **Addr 1** form the address pointer to the memory map (where necessary; may be zero if not needed).
- **Data 0** to **Data 11** are the bytes of data to be written (in the case of a write). Note that data locations beyond the number specified by NumWx will be ignored.

The response packet has the generic format given in [Figure 11-2](#).

Figure 11-2. Response Packet Format



In [Figure 11-2](#):

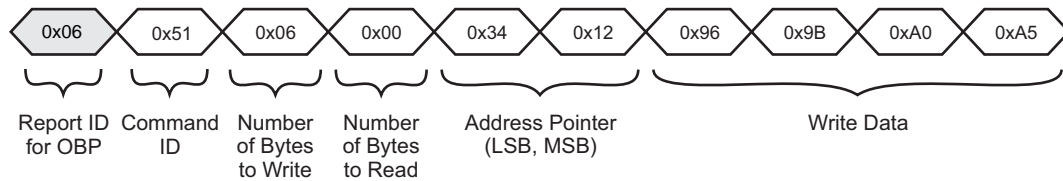
- **Rpt ID** is the HID-I²C Report ID used for Object Protocol commands and responses (see [Table 11-1 on page 36](#)).
- **Status** indicates the result of the command:
 - 0x00 = read and write completed; read data returned
 - 0x04 = write completed; no read data requested
- **NumRx** is the number of bytes following that have been read from the memory map (in the case of a read). This will be the same value as NumRx in the command packet.
- **Data 0** to **Data 14** are the data bytes read from the memory map.

11.5.1.2 Writing To the Device

A write operation cycle to the device consists of sending a packet that contains six header bytes. These specify the HID-I²C report ID, the Command ID, the number of bytes to read, the number of bytes to write, and the 16-bit address pointer. Subsequent bytes in a multi-byte transfer form the actual data. These are written to the location of the address pointer, location of the address pointer +1, location of the address pointer + 2, and so on.

Figure 11-3 shows an example command packet to write four bytes of data to contiguous addresses starting at 0x1234.

Figure 11-3. Example of a Four-byte Write Starting at Address 0x1234

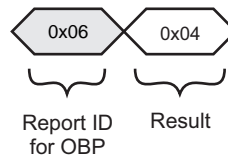


In Figure 11-3:

- **Rpt ID** is the HID-I²C Report ID used for Object Protocol commands and responses (see Table 11-1 on page 36).
- The number of bytes to read is set to zero as this is a write-only operation.
- The number of bytes to write is six: that is, four data bytes plus the two address pointer bytes.

Figure 11-4 shows the response to this command. Note that the result status returned is 0x04 (that is, the write operation was completed but no read data was requested). Note also that the Report ID is the same one used in the command packet.

Figure 11-4. Response to Example Four-byte Write

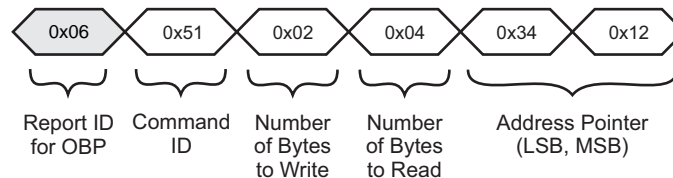


11.5.1.3 Reading From the Device

A read operation consists of sending a packet that contains the six header bytes only and no write data.

Figure 11-5 shows an example command packet to read four bytes starting at address 0x1234. Note that the address pointer is included in the number of bytes to write, so the number of bytes to write is set to 2 as there are no other data bytes to be written.

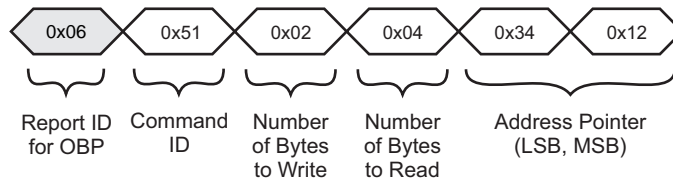
Figure 11-5. Example of a Four-byte Read Starting at Address 0x1234



It is not necessary to set the address pointer before every read. The address pointer is updated automatically after every read operation, so the address pointer will be correct if the reads occur in order.

Figure 11-6 shows the response to this command. The result status returned is 0x00 (that is the write operation was completed and the data was returned). The number of bytes returned will be the same as the number requested (4 in this case).

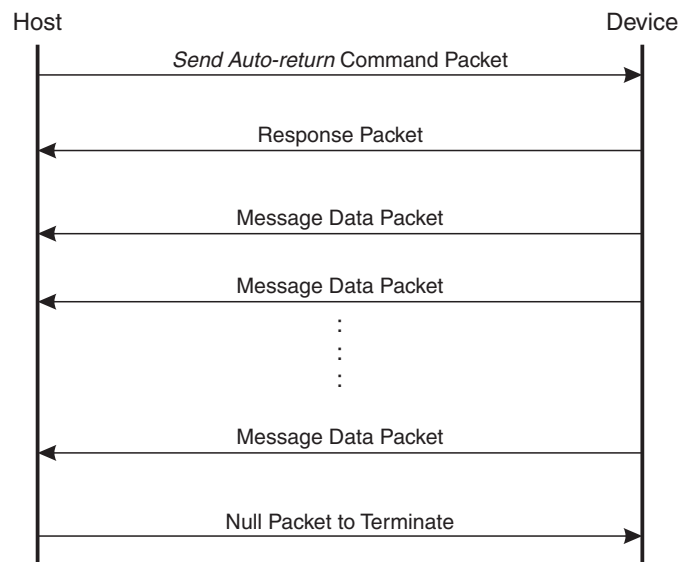
Figure 11-6. Response to Example Four-byte Read



11.5.2 Send Auto-return Messages

With this command the device can be configured to return new messages from the Message Processor T5 object autonomously. The packet sequence to do this is shown in Figure 11-7.

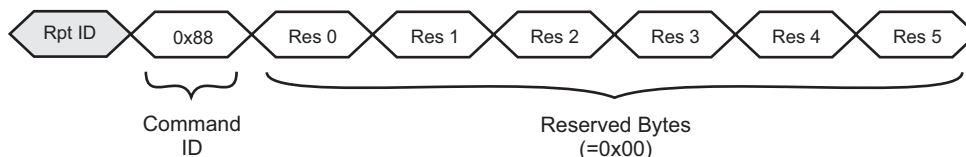
Figure 11-7. Packet Sequence for “Send Auto-return” Command



The HID-I²C Report ID used is that for Object Protocol commands and responses; see Table 11-1 on page 36 for the value.

The command packet has the format given in Figure 11-8.

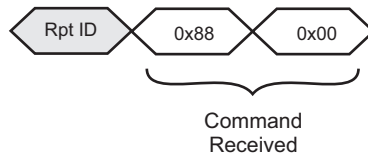
Figure 11-8. Command Packet Format



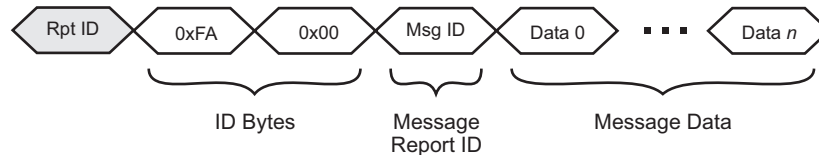
In Figure 11-8:

- **Rpt ID** is the HID-I²C Report ID used for Object Protocol commands and responses (see Table 11-1 on page 36).
- **Res 0 to Res 5** are reserved bytes with a value of 0x00.

The response packet has the format given in Figure 11-9. Note that with this command, the command packet does not include an address pointer as the device already knows the address of the Message Processor T5 object.

Figure 11-9. Response Packet Format

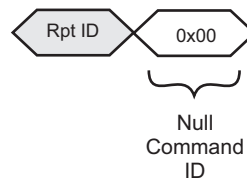
Once the device has responded to the command, it starts sending message data. Each time a message is generated in the Message Processor T5 object, the device automatically sends a message packet to the host with the data. The message packets have the format given in [Figure 11-10](#).

Figure 11-10. Message Packet Format

In [Figure 11-10](#):

- **Rpt ID** is the HID-I²C Report ID used for Object Protocol commands and responses (see [Table 11-1 on page 36](#)).
- **ID Bytes** identify the packet as an auto-return message packet.
- **Msg ID** is the Report ID returned by the Message Processor T5 object.
- **Message Data** bytes are the bytes of data returned by the Message Processor T5 object. The size of the data depends on the source object for which this is the message data. Refer to the *mXT641T 1.0 Protocol Guide* for more information.

To stop the sending of the messages, the host can send a null command packet. This consists of two bytes: a HID-I²C report ID and a command byte of 0x00 (see [Figure 11-11](#)).

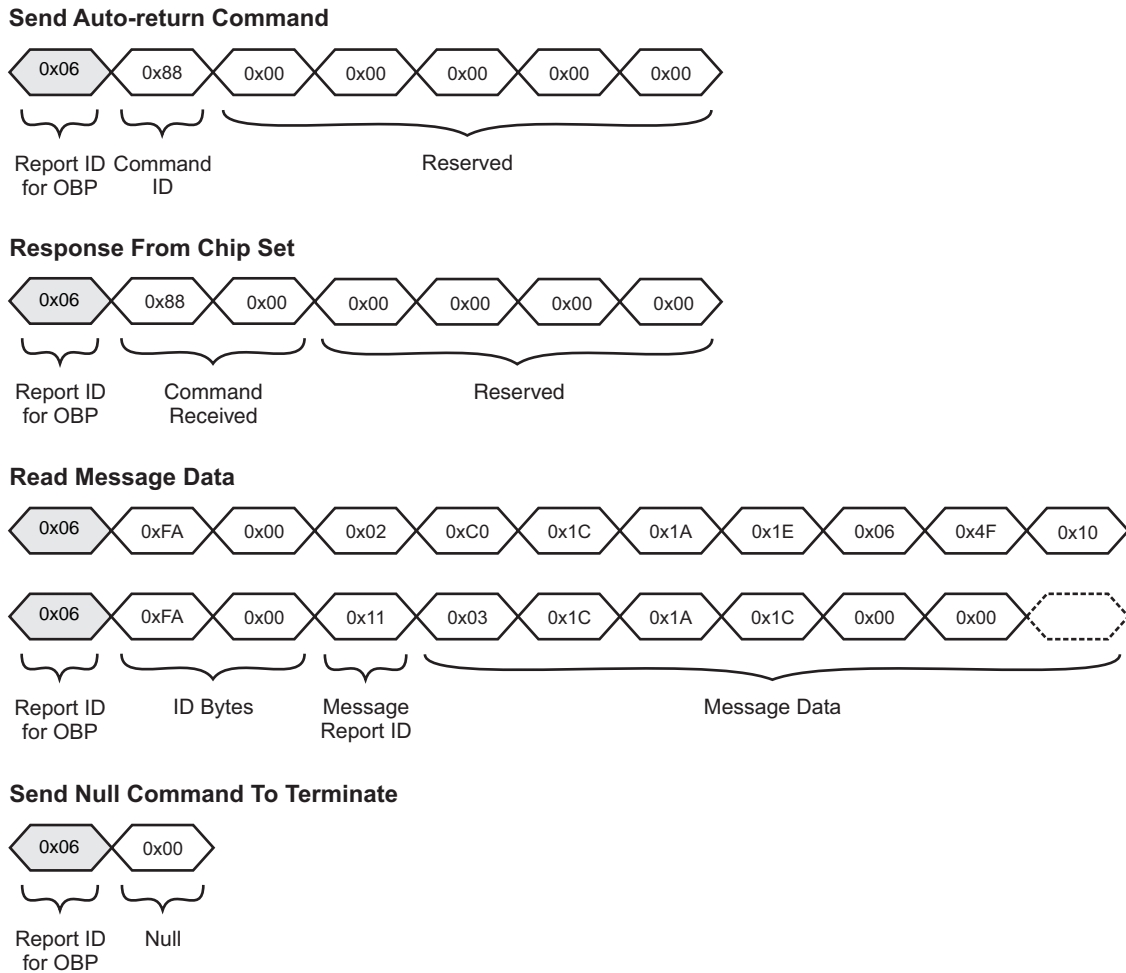
Figure 11-11. Null Command Packet Format

Note that any read or write will also terminate any currently enabled auto-return mode (see [Section 11.5.1.2 on page 38](#)).

11.5.2.1 Reading Status Messages

Figure 11-12 shows an example sequence of packets to receive messages from the Message Processor T5 object using the “Send Auto-return” command.

Figure 11-12. Example Auto-return Command Packet



11.6 Digitizer HID-I²C

This is a digitizer class HID.

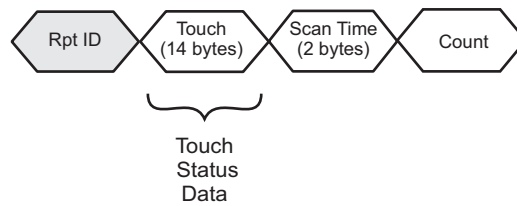
11.6.1 Touch Report

The format of a Touch report is shown in Figure 11.6.2 on page 42.

Each Touch report starts with a report ID and contains the data for one touch.

The HID-I²C Report ID used is that for Touch reports; see [Table 11-1 on page 36](#) for the value.

11.6.2 Touch Report Packet Format

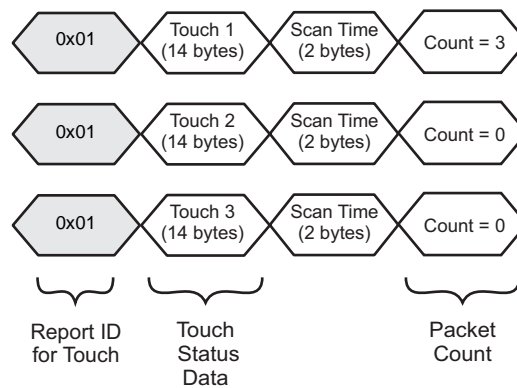


In [Figure 11.6.2](#):

- **Rpt ID** is the HID-I²C Report ID used for Touch reports (see [Table 11-1 on page 36](#)).
- **Touch** is the data for the touch.
- **Scan Time** is the Timestamp for the report packet
- **Count** is used to identify the report packets for current active touches that are to be reported as a single package. The Count in the first packet for the first touch is set to the number of active touches to be sent in one package. Subsequent packets for subsequent active touches have a Count of 0.

An example of the Touch report packets for 3 active touches is shown in [Figure 11-13](#).

Figure 11-13. Example Touch Report Packets for 3 Active Touches



Each input report consists of a HID-I²C report ID followed by 17 bytes of that describe the status of one active touch. The input report format depends on the geometry calculation control (TCHGEOMEN) of the Digitizer HID Configuration T43 object. [Table 11-2 on page 43](#) and [Table 11-3 on page 43](#) give the detailed format of a touch report packet.

Table 11-2. Touch Report Format when TCHGEOMEN = 1

Byte	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	HID-I ² C Touch Report ID							
1	Reserved							Status
2	Touch ID							
3 – 4	X Position							
5 – 6	X' Position							
7 – 8	Y Position							
9 – 10	Y' Position							
11	Touch Width							
12	Reserved							
13	Touch Height							
14	Reserved							
15 – 16	Scan Time							
17	Count							

Table 11-3. Touch Report Format when TCHGEOMEN = 0

Byte	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
0	HID-I ² C Touch Report ID							
1	Reserved							Status
2	Touch ID							
3 – 4	X Position							
5 – 6	X' Position							
7 – 8	Y Position							
9 – 10	Y' Position							
11	Reserved							
12	Reserved							
13	Reserved							
14	Reserved							
15 – 16	Scan Time							
17	Count							

- **Byte 0:**
The HID-I²C Report ID (see [Table 11-1 on page 36](#) for Touch reports).
- **Byte 1:**
Status: Status of the touch detection. This bit is set to 1 if touch is detected, and set to 0, if no touches are detected.
- **Byte 2:**
Touch ID: Identifies the touch for which this is a status report (starting from 0).
- **Bytes 3 to 10:**

X and Y positions: These are scaled to 12-bit resolution. This means that the upper four bits of the MSByte will always be zero.

- **Byte 11:**

Touch Width: Reports the width of the detected touch when TCHGEOMEN is set to 1.

Reserved when TCHGEOMEN is set to 0

- **Byte 13:**

Touch Height: Reports the height of the detected touch when TCHGEOMEN is set to 1.

Reserved when TCHGEOMEN is set to 0

- **Byte 15 to 16:**

Scan Time: Timestamp associated with the current report packet with a 10 kHz resolution.

- **Byte 17:**

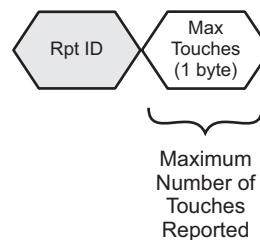
Count: For the first touch, the number of active touches to be sent in one package. Subsequent packets for subsequent active touches have a Count of 0.

11.6.3 Maximum Touches (Surface Contacts) Report

The format of the Maximum Touches report packet is shown in [Figure 11-14](#).

The HID-I²C Report ID used is that for Maximum Touches reports; see [Table 11-1 on page 36](#) for the value.

Figure 11-14. Example Maximum Touches Report



In [Figure 11-14](#):

- **Rpt ID** is the HID-I²C Report ID used for Maximum Touches reports (see [Table 11-1 on page 36](#)).
- **Max Touches** is the maximum number of touches that can be reported by the device.

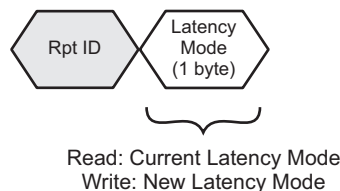
Read this report to receive the maximum number of touches that can currently be reported.

Write this report to set the maximum number of touches to be reported. Note that the number of touches cannot be set to more than the maximum number of touches defined by Multiple Touch Touchscreen T100 NUMTCH.

11.6.4 Touch Hardware Quality Assurance (THQA) Report

The THQA data is reported to Windows using the THQA Report ID (see [Table 11-1 on page 36](#) for the value). The content of this data is defined by Microsoft.

11.7 $\overline{\text{CHG}}$ Line



The $\overline{\text{CHG}}$ line is an active-low, open-drain output that is used to alert the host that a new message is available in the Input Buffer. This provides the host with an interrupt-style interface with the potential for fast response times. It reduces the need for wasteful I²C communications.

Further information on the $\overline{\text{CHG}}$ line is given in [Section 10.6 on page 33](#).

11.8 SDA, SCL

Identical to standard I²C operation. Refer to [Section 10.7 on page 34](#).

11.9 Clock Stretching

Identical to standard I²C operation. Refer to [Section 10.8 on page 35](#).

11.10 Power Control

The mXT641T supports the use of the HID-I²C *SET POWER* commands to put the device into a low power state.

11.11 Microsoft Windows Compliance

The mXT641T has algorithms within the Digitizer HID Configuration T43 and Multiple Touch Touchscreen T100 specifically to ensure Microsoft Windows 8 and 8.1 compliance.

The device also supports Microsoft Touch Hardware Quality Assurance (THQA) in the Serial Data Command T68 object. Refer to the Microsoft whitepaper *How to Design and Test Multitouch Hardware Solutions for Windows 8*.

These, and other device features, may need specific tuning.

12. PCB Design Considerations

12.1 Introduction

The following sections give the design considerations that should be adhered to when designing a PCB layout for use with the mXT641T. Of these, power supply and ground tracking considerations are the most critical.

By observing the following design rules, and with careful preparation for the PCB layout exercise, designers will be assured of a far better chance of success and a correctly functioning product.

12.2 Printed Circuit Board

Atmel recommends the use of a four-layer printed circuit board for mXT641T applications. This, together with careful layout, will ensure that the board meets relevant EMC requirements for both noise radiation and susceptibility, as laid down by the various national and international standards agencies.

12.2.1 PCB Cleanliness

Modern no-clean-flux is generally compatible with capacitive sensing circuits.



CAUTION: If a PCB is reworked to correct soldering faults relating to any of the device devices, or to any associated traces or components, be sure that you fully understand the nature of the flux used during the rework process. Leakage currents from hygroscopic ionic residues can stop capacitive sensors from functioning. If you have any doubts, a thorough cleaning after rework may be the only safe option.

12.3 Supply Rails and Ground Tracking

Power supply and clock distribution are the most critical parts of any board layout. Because of this, it is advisable that these be completed before any other tracking is undertaken. After these, supply decoupling, and analog and high speed digital signals should be addressed. Track widths for all signals, especially power rails should be kept as wide as possible in order to reduce inductance.

The Power and Ground planes themselves can form a useful capacitor. Flood filling for either or both of these supply rails, therefore, should be used where possible. It is important to ensure that there are no floating copper areas remaining on the board: all such areas should be connected to the 0 V plane. The flood filling should be done on the outside layers of the board.

12.4 Power Supply Decoupling

As a rule, a suitable decoupling capacitor should be placed on each and every supply pin on all digital devices. It is important that these capacitors are placed as close to the chip supply pins as possible (less than 10 mm away). The ground connection of these capacitors should be tracked to 0 V by the shortest, heaviest traces possible.

Capacitors with a Type II dielectric, such as X5R or X7R and with a value of at least 100 nF, should be used for this purpose.

In addition, at least one 'bulk' decoupling capacitor, with a minimum value of 4.7 μ F should be placed on each power rail, close to where the supply enters the board.

Surface mounting capacitors are preferred to wire leaded types due to their lower ESR and ESL. It is often possible to fit these decoupling capacitors underneath and on the opposite side of the PCB to the digital ICs. This will provide the shortest tracking, and most effective decoupling possible.

Refer to the application note *Selecting Decoupling Capacitors for Atmel PLDs* (doc0484.pdf; available on the Atmel website) for further general information on decoupling capacitors.

12.5 Single Supply Operation

When designing a PCB for an application using a single LDO, extra care should be taken to ensure short, low inductance traces between the supply and the touch controller supply input pins. Ideally, tracking for the individual supplies should be arranged in a star configuration, with the LDO at the junction of the star. This will ensure that supply current variations or noise in one supply rail will have minimum effect on the other supplies. In applications where a ground plane is not practical, this same star layout should also apply to the power supply ground returns.

12.6 Analog I/O

In general, tracking for the analog I/O signals from the device should be kept as short as possible. These normally go to a connector which interfaces directly to the touchscreen.

Ensure that adequate ground-planes are used. An analog ground plane should be used in addition to a digital one. Care should be taken to ensure that both ground planes are kept separate and are connected together only at the point of entry for the power to the PCB. This is usually at the input connector.

12.7 Component Placement and Tracking

It is important to orient all devices so that the tracking for important signals (such as power and clocks) are kept as short as possible. This simple point is often overlooked when initially planning a PCB layout and can save hours of work at a later stage.

12.7.1 Digital Signals

In general, when tracking digital signals, it is advisable to avoid sharp directional changes, sensitive signal tracks (such as analog I/O) and any clock or crystal tracking.

A good ground return path for all signals should be provided, where possible, to ensure that there are no discontinuities in the ground return path.

12.8 EMC and Other Observations

The following recommendations are not mandatory, but may help in situations where particularly difficult EMC or other problems are present:

- Try to keep as many signals as possible on the inside layers of the board. If suitable ground flood fills are used on the top and bottom layers, these will provide a good level of screening for noisy signals, both into and out of the PCB.
- Ensure that the on-board regulators have sufficient tracking around and underneath the devices to act as a heatsink. This heatsink will normally be connected to the 0 V or ground supply pin. Increasing the width of the copper tracking to any of the device pins will aid in removing heat. There should be no solder mask over the copper track underneath the body of the regulators.
- Ensure that the decoupling capacitors, especially high capacity ceramic type, have the requisite low ESR, ESL and good stability/temperature properties. Refer to the regulator manufacturer's datasheet for more information.

13. Getting Started with mXT641T

13.1 Establishing Contact

13.1.1 Communication with the Host

The host can use the following interface to communicate with the device:

- I²C interface (see [Section 10. on page 30](#))
- HID-I²C interface (see [Section 11. on page 36](#))

13.1.2 Power-up Sequence

On power-up, the $\overline{\text{CHG}}$ line goes low to indicate that there is new data to be read from the Message Processor T5 object. If the $\overline{\text{CHG}}$ line does not go low, there is a problem with the device.

The host should attempt to read any available messages to establish that the device is present and running following power-up or a reset. Examples of messages include reset or calibration messages. The host should also check that there are no configuration errors reported.

13.2 Using the Object Protocol

The device has an object-based protocol that is used to communicate with the device. Typical communication includes configuring the device, sending commands to the device, and receiving messages from the device. Refer to the *mXT641T 1.0 Protocol Guide* for more information.

The host must perform the following initialization so that it can communicate with the device:

1. Read the start positions of all the objects in the device from the Object Table and build up a list of these addresses.
2. Use the Object Table to calculate the report IDs so that messages from the device can be correctly interpreted.

13.3 Writing to the Device

There are a number of mechanisms for writing to the device:

- Using an I²C write operation (see [Section 10.2 on page 30](#)).
- Using the Generic HID-I²C write operation (see [Section 11.5.1.2 on page 38](#)).

To communicate with the device, you write to the appropriate object:

- To send a command to the device, you write the appropriate command to the Command Processor T6 object (refer to the *mXT641T 1.0 Protocol Guide*).
- To configure the device, you write to an object. For example, to configure the device power consumption you write to the global Power Configuration T7 object, and to set up a touchscreen you write to a Multiple Touch Touchscreen T100 object. Some objects are optional and need to be enabled before use. Refer to the *mXT641T 1.0 Protocol Guide* for more information on the objects.

13.4 Reading from the Device

Status information is stored in the Message Processor T5 object. This object can be read to receive any status information from the device. The following mechanisms provide an interrupt-style interface for reading messages in the Message Processor T5 object:

- The $\overline{\text{CHG}}$ line is asserted whenever a new message is available in the Message Processor T5 object (see [Section 10.6 on page 33](#)). See [Section 10.4 on page 31](#) for information on the format of the I²C read operation.
- When using the HID-I²C interface, the interface provides an interrupt-driven interface that sends the messages automatically (see [Section 11.5.1.3 on page 38](#))

Note that the host should always wait to be notified of messages. The host should not poll the device for messages.

13.5 Configuring the Device

The objects are designed such that a default value of zero in their fields is a “safe” value that typically disables functionality. The objects must be configured before use and the settings written to the nonvolatile memory using the Command Processor T6 object.

Perform the following actions for each object:

1. Enable the object, if the object requires it.
2. Configure the fields in the object, as required.
3. Enable reporting, if the object supports messages, to receive messages from the object.

Refer to the *mXT641T 1.0 Protocol Guide* for more information on configuring the objects.

The following objects require no configuration:

- Debug Objects
 - Diagnostic Debug T37
- General objects:
 - Message Processor T5
 - Command Processor T6
- Support objects:
 - Message Count T44

The following objects must be configured before use:

- General objects
 - Power Configuration T7
 - Acquisition Configuration T8

The following objects should be checked and configured as necessary:

- Touch objects:
 - Key Array T15
 - Multiple Touch Touchscreen T100
- Signal processing objects:
 - Stylus T47
- Support objects:
 - Communications Configuration T18
 - GPIO/PWM Configuration T19
 - User Data T38
 - Digitizer HID Configuration T43
 - CTE Configuration T46
 - Touchscreen Hover Configuration T101
 - Self Capacitance Global Configuration T109
 - Self Capacitance Tuning Parameters T110
 - Self Capacitance Configuration T111
 - Self Capacitance Measurement Configuration T113

The following objects are optional and can be configured, as required:

- Signal processing objects:
 - Grip Suppression T40
 - Touch Suppression T42
 - Shieldless T56
 - Lens Bending T65
 - maXCharger T72
 - Glove Detection T78

- Retransmission Compensation T80
- Unlock Gesture T81
- Gesture Processor T84
- Symbol Gesture Processor T92
- Touch Sequence Processor T93
- Self Capacitance maXCharger T108
- Self Capacitance Grip Suppression T112
- Support objects:
 - Self Test T25
 - Timer T61
 - Serial Data Command T68
 - Dynamic Configuration Controller T70
 - Dynamic Configuration Container T71
 - CTE Scan Configuration T77
 - Auxiliary Touch Configuration T104

14. Debugging

The device provides a mechanism for obtaining raw data for development and testing purposes by reading data from the Diagnostic Debug T37 object. Refer to the *mXT641T 1.0 Protocol Guide* for more information on this object.

A second mechanism is provided that allows the host to read the real-time raw data using the low-level debug port. This can be accessed via the SPI interface. Refer to QTAN0050, *Using the maXTouch Debug Port*, for more information on the debug port.

There is also a Self Test T25 object that runs self-test routines in the device to find hardware faults on the sense lines and the electrodes. This object also performs an initial pin fault test on power-up to ensure that there is no X-to-Y short before the high-voltage supply is enabled inside the chip. A high-voltage short into the analog circuitry would break the device.

Refer to the *mXT641T 1.0 Protocol Guide* and QTAN0059, *Using the maXTouch Self Test Feature*, for more information on the Self Test T25 object.

15. Specifications

15.1 Absolute Maximum Specifications

Vdd	3.6 V
VddIO	3.6 V
AVdd	3.6 V
XVdd	9.0 V
Voltage forced onto any pin	-0.3 V to (Vdd, VddIO or AVdd) + 0.3 V
Configuration parameters maximum writes (flash memory write cycles)	10,000



CAUTION: Stresses beyond those listed under *Absolute Maximum Specifications* may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum specification conditions for extended periods may affect device reliability.

15.2 Recommended Operating Conditions

Operating temp	-40°C to +85°C
Storage temp	-60°C to +150°C
Vdd	3.3 V
VddIO	1.8 V to 3.3 V
AVdd	3.3 V
XVdd with internal voltage doubler	Vdd to 2 × Vdd
Cx transverse load capacitance per node	0.6 pF to 3 pF
Temperature slew rate	10°C/min

15.2.1 DC Characteristics

15.2.1.1 Analog Voltage Supply – AVdd

Parameter	Min	Typ	Max	Units	Notes
Operating limits	2.7	3.3	3.47	V	
Supply Rise Rates	–	–	0.25	V/ μ s	

15.2.1.2 Digital Voltage Supply

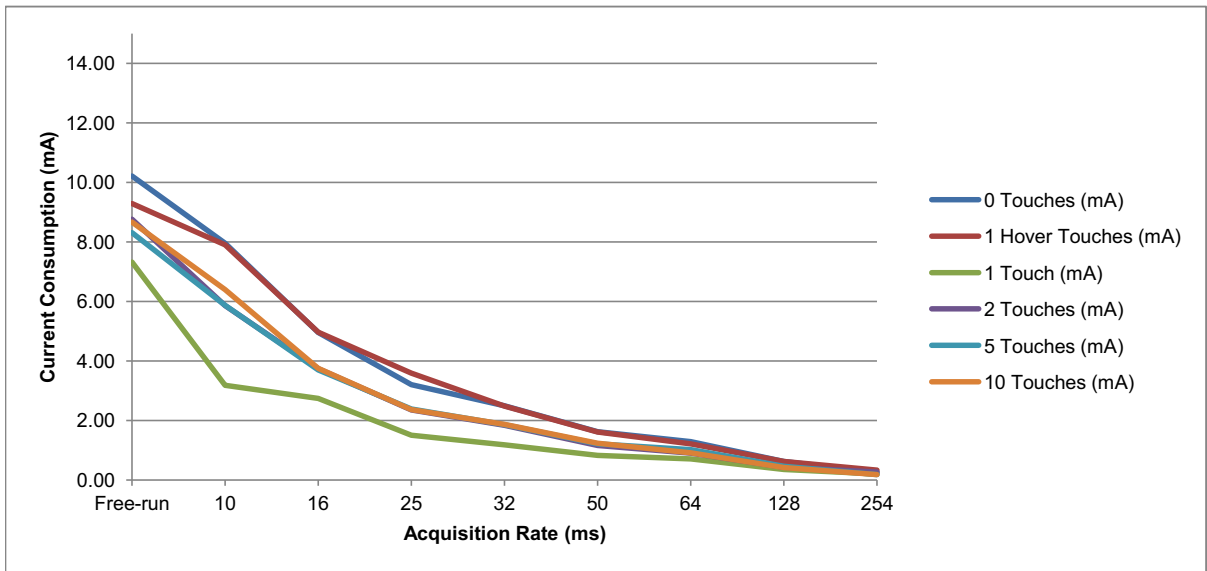
Parameter	Min	Typ	Max	Units	Notes
VddIO					
Operating limits	1.71	–	3.47	V	I ² C-compatible
Supply Rise Rates	–	–	0.25	V/ μ s	
Vdd					
Operating limits	3.0	3.3	3.47	V	
Supply Rise Rates	–	–	0.25	V/ μ s	

15.3 Supply Current

15.3.1 Analog Supply

I2C Interface, AVdd = 3.294V

Acquisition Rate (ms)	0 Touches (mA)	1 Hover Touches (mA)	1 Touch (mA)	2 Touches (mA)	5 Touches (mA)	10 Touches (mA)
Free-run	10.21	9.29	7.32	8.76	8.30	8.66
10	7.96	7.90	3.18	5.86	5.87	6.40
16	4.95	4.97	2.74	3.74	3.70	3.75
25	3.20	3.58	1.50	2.35	2.38	2.36
32	2.49	2.48	1.18	1.83	1.85	1.86
50	1.62	1.61	0.82	1.16	1.21	1.23
64	1.29	1.21	0.71	0.90	1.02	0.91
128	0.62	0.62	0.35	0.42	0.46	0.42
254	0.25	0.33	0.20	0.26	0.18	0.18

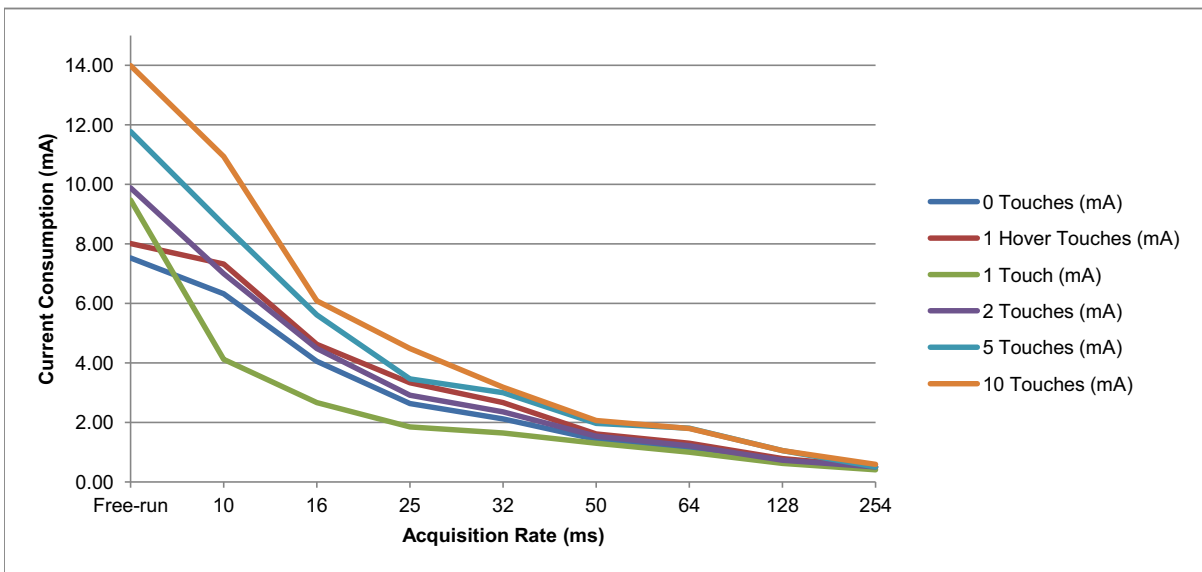


15.3.2 Digital Supply

15.3.2.1 Vdd

I2C Interface, Vdd = 3.281V

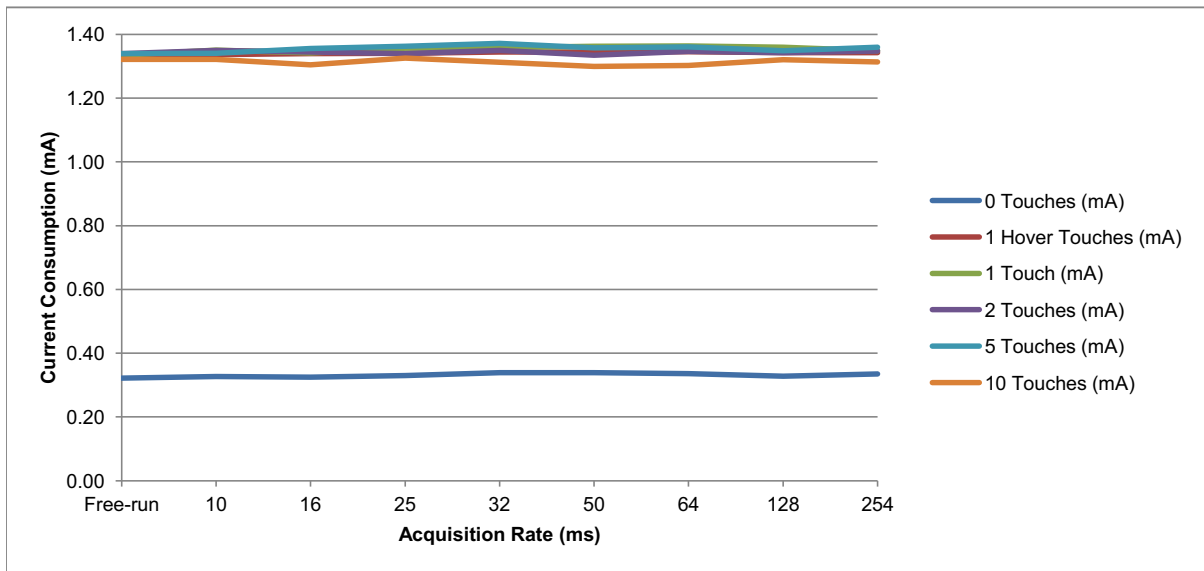
Acquisition Rate (ms)	0 Touches (mA)	1 Hover Touches (mA)	1 Touch (mA)	2 Touches (mA)	5 Touches (mA)	10 Touches (mA)
Free-run	7.52	8.01	9.47	9.88	11.78	13.98
10	6.32	7.33	4.11	7.00	8.64	10.94
16	4.05	4.62	2.67	4.48	5.61	6.09
25	2.64	3.33	1.84	2.91	3.46	4.48
32	2.11	2.66	1.64	2.35	3.00	3.18
50	1.42	1.61	1.30	1.54	1.97	2.06
64	1.10	1.30	0.99	1.22	1.81	1.79
128	0.72	0.78	0.62	0.73	1.05	1.06
254	0.48	0.50	0.40	0.50	0.50	0.59



15.3.3 VddIO Current

I2C Interface, VddIO = 3.294V

Acquisition Rate (ms)	0 Touches (mA)	1 Hover Touches (mA)	1 Touch (mA)	2 Touches (mA)	5 Touches (mA)	10 Touches (mA)
Free-run	0.32	1.34	1.33	1.34	1.34	1.32
10	0.33	1.34	1.35	1.35	1.34	1.32
16	0.33	1.34	1.34	1.34	1.36	1.30
25	0.33	1.34	1.36	1.34	1.36	1.33
32	0.34	1.34	1.36	1.35	1.37	1.31
50	0.34	1.34	1.36	1.33	1.36	1.30
64	0.34	1.35	1.36	1.35	1.36	1.30
128	0.33	1.34	1.36	1.34	1.35	1.32
254	0.33	1.34	1.35	1.35	1.36	1.31



15.4 Sleep Current

Parameter	I ² C	Units	Notes
A _{Idd}	0.24	μA	AV _{dd} = 3.287 V
D _{Idd}	149.2	μA	V _{dd} = 3.3 V
D _{IddIO}	332.3	μA	V _{dd} = 3.3 V

15.5 Deep Sleep Current

$T_A = 25^\circ\text{C}$

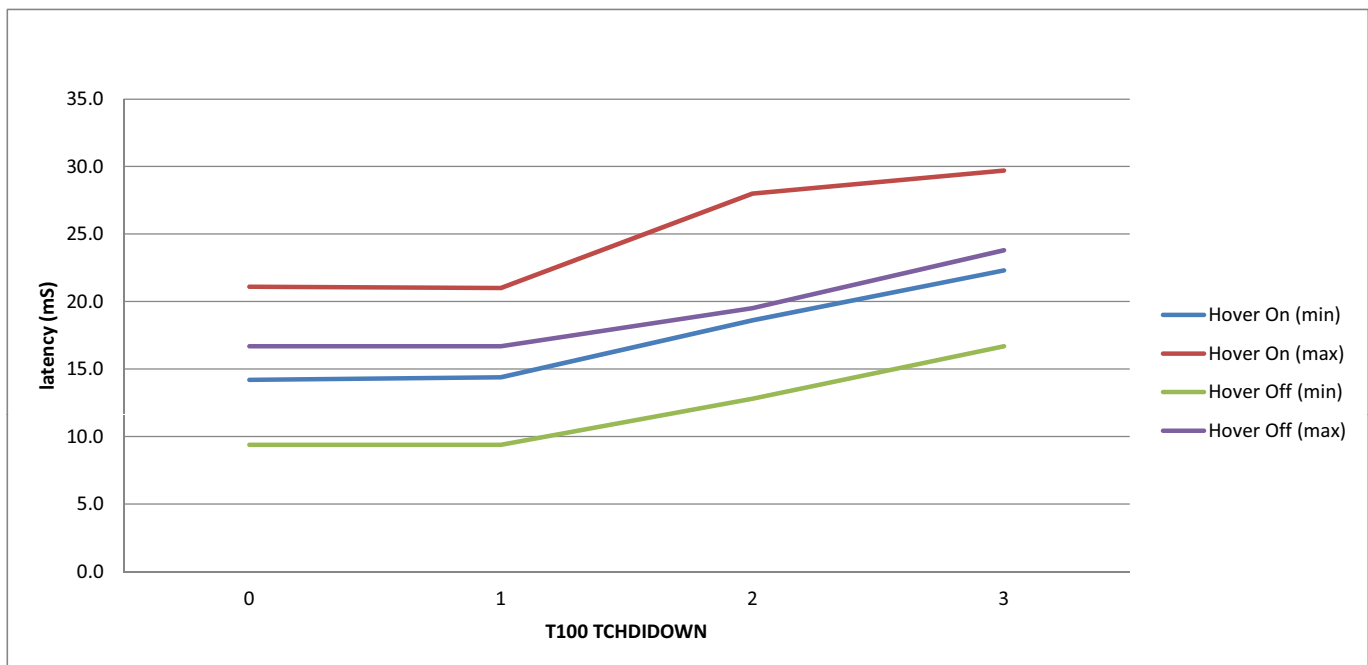
Parameter	Min	Typ	Max	Units	Notes
Deep Sleep Current	–	105	–	μA	Vdd = 3.3 V, AVdd = 3.3 V
Deep Sleep Power	–	346	–	μW	Vdd = 3.3 V, AVdd = 3.3 V

15.6 Power Supply Ripple and Noise

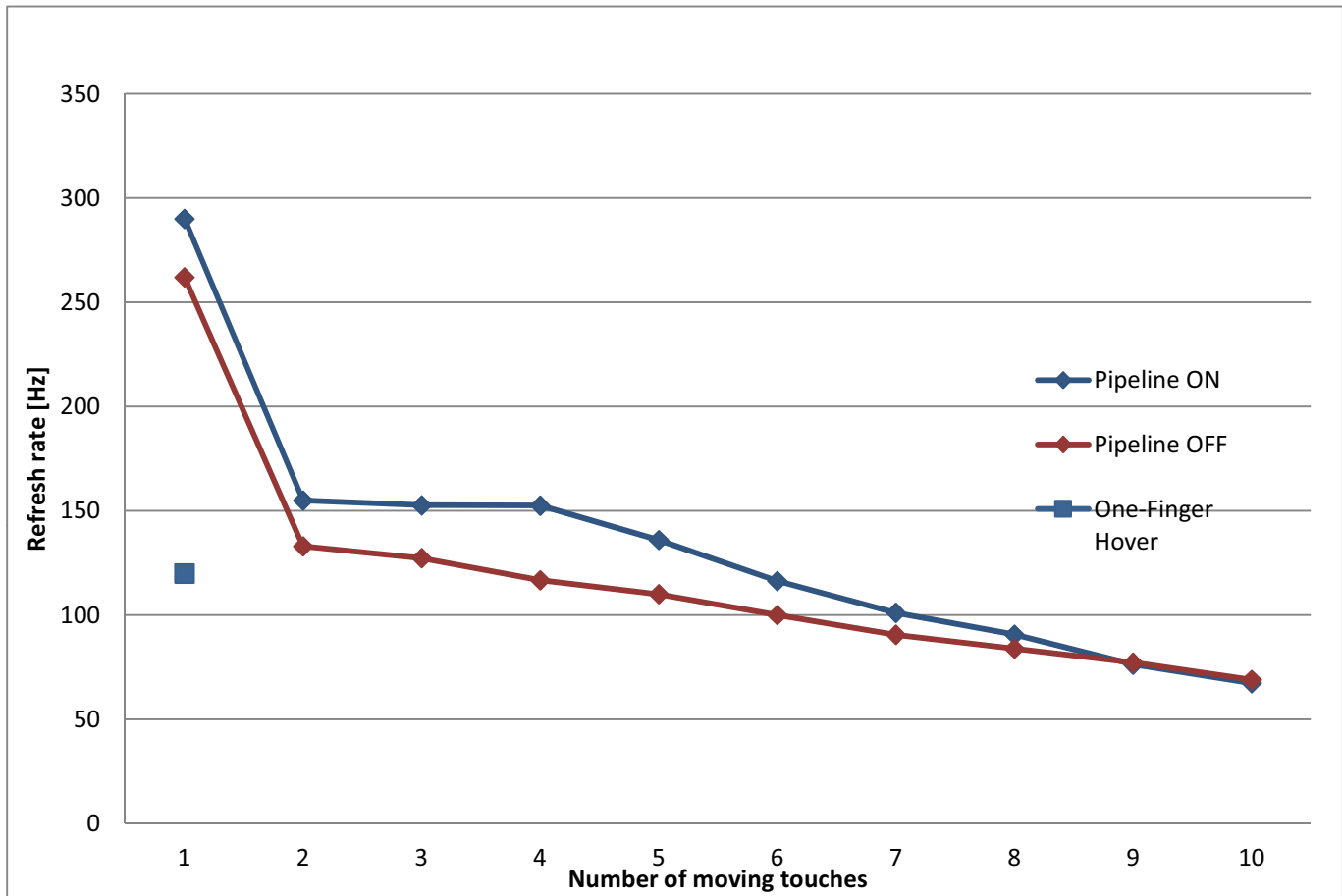
Parameter	Min	Typ	Max	Units	Notes
Vdd	–	–	± 50	mV	Across frequency range 1 Hz to 1 MHz
AVdd	–	–	± 25	mV	Across frequency range 1 Hz to 1 MHz
AVdd (with noise suppression enabled)	–	–	± 40	mV	Across frequency range 1 Hz to 1 MHz, with Noise Suppression enabled

15.7 Timing Specifications

15.7.1 Touch Latency



15.7.2 Speed



15.7.3 Reset Timing

Parameter	Min	Typ	Max	Units	Notes
Power on to $\overline{\text{CHG}}$ line low	–	79	–	ms	Vdd supply for POR VddIO supply for external reset
Hardware reset to $\overline{\text{CHG}}$ line low	–	79	–	ms	
Software reset to $\overline{\text{CHG}}$ line low	–	95	–	ms	

The mXT641T meets Microsoft Windows 8 requirements.

Note: Any $\overline{\text{CHG}}$ line activity before the power-on or reset period has expired should be ignored by the host. Operation of this signal cannot be guaranteed before the power-on/reset periods have expired (see [Table 15.7.3 on page 58](#)).

15.8 Input/Output Characteristics

Parameter	Description	Min	Typ	Max	Units	Notes
Input ($\overline{\text{RESET}}$, GPIO, SDA, SCL)						
Vil	Low input logic level	-0.3	-	$0.3 \times V_{\text{ddIO}}$	V	$V_{\text{ddIO}} = 1.8 \text{ V to Vdd}$
Vih	High input logic level	$0.6 \times V_{\text{ddIO}}$	-	V_{ddIO}	V	$V_{\text{ddIO}} = 1.8 \text{ V to Vdd}$
Iil	Input leakage current	-	-	1	μA	Pull-up resistors disabled
$\overline{\text{RESET}}$ pin	Internal pull-up resistor	20	40	60	$\text{k}\Omega$	
Output (CHG, GPIO)						
Vol	Low output voltage	0	-	$0.2 \times V_{\text{ddIO}}$	V	$V_{\text{DDIO}} = 1.8 \text{ V to VDD}$. Iol = -2mA
Voh	High output voltage	$0.8 \times V_{\text{ddIO}}$	-	V_{ddIO}	V	$V_{\text{DDIO}} = 1.8 \text{ V to VDD}$. Ioh = 2mA

15.9 I²C Specifications

Parameter	Value
Addresses	0x4A or 0x4B
Maximum bus speed (SCL)	3.4 MHz
I ² C specification	Version 6.0
Required pull-up resistance for standard mode (100 kHz)	1 k Ω to 10 k Ω ⁽¹⁾
Required pull-up resistance for fast mode (400 kHz)	1 k Ω to 3 k Ω ⁽¹⁾
Required pull-up resistance for fast+ mode (1 MHz)	0.7 k Ω (max) ⁽¹⁾
Required pull-up resistance for high-speed mode (3.4 MHz)	0.5 k Ω to 0.75 k Ω ⁽¹⁾

Notes: 1. The values of pull-up resistors should be chosen to ensure SCL and SDA rise and fall times meet the I²C specification. The value required will depend on the amount of stray capacitance on the line.

15.10 HID-I²C Specification

Parameter	Value
Vendor ID	0x03EB (Atmel)
Product ID	0x214D (mXT641T)
HID-I ² C specification	1.0

15.11 Touch Accuracy and Repeatability

Parameter	Min	Typ	Max	Units	Notes
Linearity (touch only; 5.4 mm electrode pitch)	–	±1	–	mm	8 mm or greater finger
Linearity (touch only; 4.2 mm electrode pitch)	–	±0.5	–	mm	4 mm or greater finger
Accuracy	–	±1	–	mm	
Accuracy at edge	–	±2	–	mm	
Repeatability	–	±0.25	–	%	X axis with 12-bit resolution

15.12 Thermal Packaging

15.12.1 Thermal Data

Parameter	Typ	Unit	Condition	Package
Junction to ambient thermal resistance	57.3	°C/W	Still air	UFBGA 84, 6 × 6 mm
Junction to case thermal resistance	8.7	°C/W		UFBGA 84, 6 × 6 mm
Junction to ambient thermal resistance	57.6	°C/W	Still air	XFLGA 84, 6 × 6 mm
Junction to case thermal resistance	8.7	°C/W		XFLGA 84, 6 × 6 mm

15.12.2 Junction Temperature

The average chip junction temperature, T_J in °C can be obtained from the following:

$$T_J = T_A + (P_D \times \theta_{JA})$$

If a cooling device is required, use this equation:

$$T_J = T_A + (P_D \times (\theta_{HEATSINK} + \theta_{JC}))$$

where:

- θ_{JA} = package thermal resistance, Junction to ambient (°C/W).
- θ_{JC} = package thermal resistance, Junction to case thermal resistance (°C/W).
- $\theta_{HEATSINK}$ = cooling device thermal resistance (°C/W), provided in the cooling device datasheet.
- P_D = device power consumption (W).
- T_A is the ambient temperature (°C).

15.13 ESD Information

Parameter	Value	Reference standard
Human Body Model (HBM)	±2000 V	JEDEC JS-001
Charge Device Model (CDM)	±250 V	

15.14 Soldering Profile

Profile Feature	Green Package
Average Ramp-up Rate (217°C to Peak)	3°C/s max
Preheat Temperature 175°C ±25°C	150 – 200°C
Time Maintained Above 217°C	60 – 150 s
Time within 5°C of Actual Peak Temperature	30 s
Peak Temperature Range	260°C
Ramp down Rate	6°C/s max
Time 25°C to Peak Temperature	8 minutes max

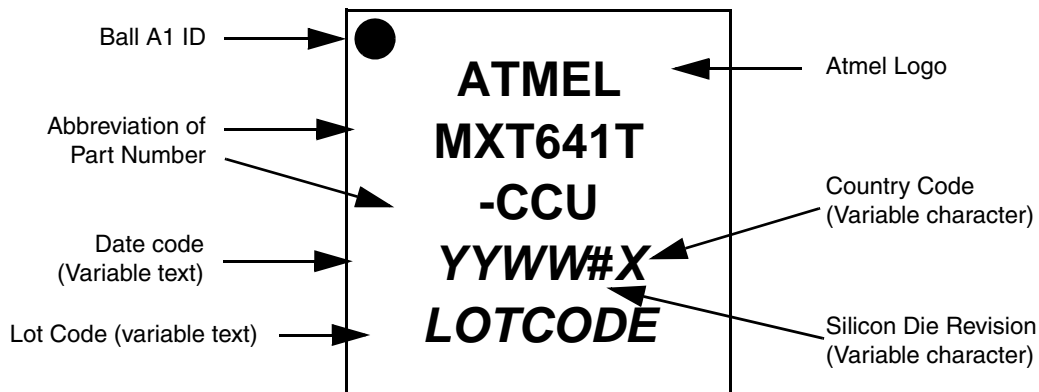
15.15 Moisture Sensitivity Level (MSL)

MSL Rating	Package Type(s)	Peak Body Temperature	Specifications
MSL3	BGA	260°C	IPC/JEDEC J-STD-020

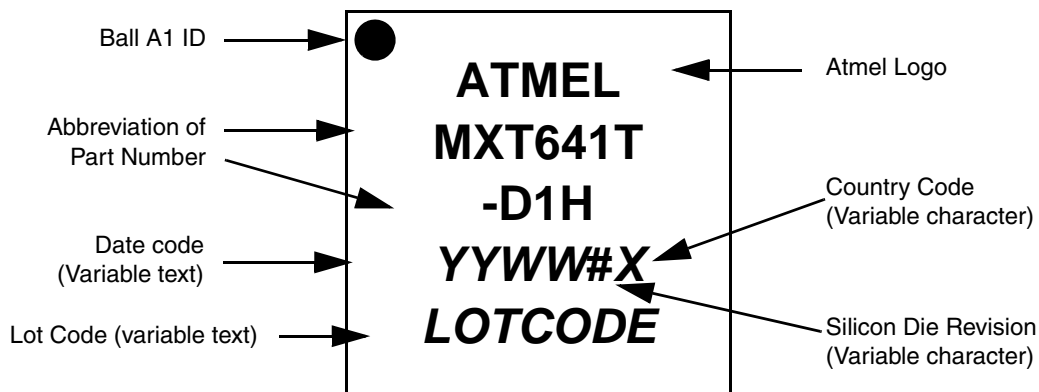
16. Package Information

16.1 Part Markings

16.1.1 ATMXT641T-CCU



16.1.2 ATMXT641T-D1H

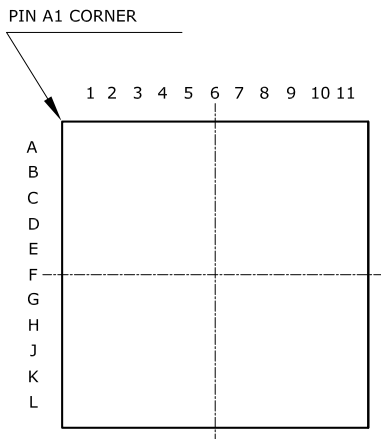


16.2 Orderable Part Numbers

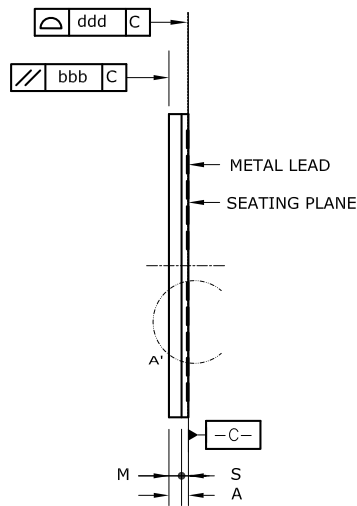
Orderable Part Number	QS Number	Firmware Revision	Description
ATMXT641T-CCU (Supplied in trays)	854	1.0	84-ball UFBGA 6 × 6 × 0.6 mm, 0.5 mm ball pitch, RoHS compliant
ATMXT641T-CCUR (Supplied in tapes and reels)	854	1.0	84-ball UFBGA 6 × 6 × 0.6 mm, 0.5 mm ball pitch, RoHS compliant
ATMXT641T-D1H (Supplied in trays)	854	1.0	84-pad XFLGA 6 × 6 × 0.5 mm, 0.5 mm pad pitch, RoHS compliant
ATMXT641T-D1HR (Supplied in tapes and reels)	854	1.0	84-pad XFLGA 6 × 6 × 0.5 mm, 0.5 mm pad pitch, RoHS compliant

16.3.2 XFLGA 84 Pads

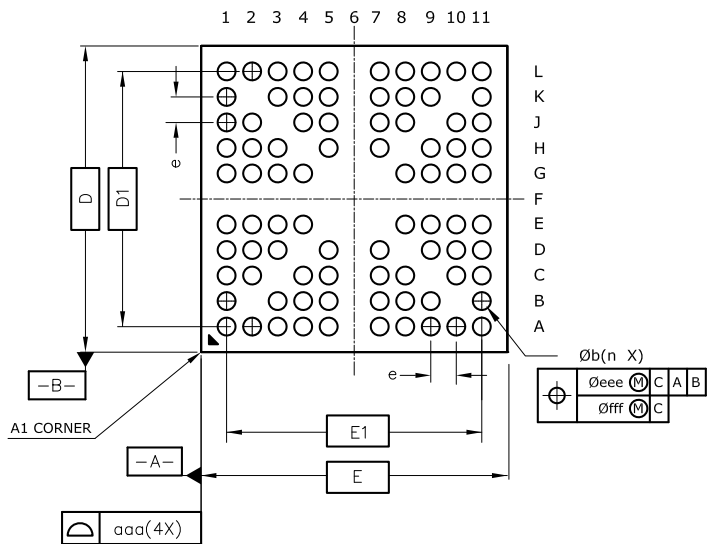
DRAWINGS NOT SCALED



TOP VIEW



SIDE VIEW



BOTTOM VIEW

COMMON DIMENSIONS

(Unit of Measure = mm)

SYMBOL	MIN	NOM	MAX	NOTE
A	-----	-----	0.500	2
E/D	6.00 / 6.00			
E1/D1	5.00 / 5.00			
b	0.200	0.250	0.300	
e	Pitch : 0.500			
M	Mold thickness : 0.250 ref			
S	Subst thickness : 0.136 ref			
aaa	Pack edge tolerance : 0.100			
bbb	Mold flatness : 0.100			
ddd	Copla : 0.080			
n	84			

- Notes : 1. No JEDEC Drawing Reference.
 2. Dimension A includes package body thickness, and lid height, but does not include attached features.

07/01/2013

<p>Package Drawing Contact: packagedrawings@atmel.com</p>	TITLE	GPC	DRAWING NO.	REV.
	AE, 84 Leads (11x11 Custom, 4P Array), 0.5mm pitch, 6x6x0.5mm Extremely Thin Fine Pitch Land Grid Array Package (XFLGA)	DFG	AE	A

Appendix A. QMatrix Primer

A.1 Acquisition Technique

QMatrix capacitive acquisition uses a series of pulses to deposit charge into a sampling capacitor, C_s . The pulses are driven on X lines from the controller. The rising edge of the pulse causes current to flow in the mutual capacitance, C_x , formed between the X line and a neighboring receiver electrode or Y line. While one X line is being pulsed, all others are grounded. This leads to excellent isolation of the particular mutual capacitances being measured⁽¹⁾, a feature that makes for good inherent touchscreen performance.

After a fixed number of pulses (known as the burst length) the sampling capacitor's voltage is measured to determine how much charge has accumulated. This charge is directly proportional to C_x and therefore changes if C_x ⁽²⁾ changes. The transmit-charge transfer process between the X lines and Y lines causes an electric field to form that loops from X to Y. The field itself emanates from X and terminates on Y. If the X and Y electrodes are fixed directly⁽³⁾ to a dielectric material like plastic or glass, then this field tends to channel through the dielectric with very little leakage of the field out into free-space (that is, above the panel). Some proportion of the field does escape the surface of the dielectric, however, and so can be influenced during a touch.

When a finger is placed in close proximity (a few millimeters) or directly onto the dielectric's surface, some of this stray field and some of the field that would otherwise have propagated via the dielectric and terminated onto the Y electrode, is diverted into the finger and is conducted back to the controller chip via the human body rather than via the Y line.

This means that less charge is accumulated in C_s , and hence the terminal voltage present on C_s , after all the charge transfer pulses are complete, becomes less. In this way, the controller can measure changes in C_x during touch. This means that the measured capacitance C_x goes down during touch, because the coupled field is partly diverted by the touching object.

The spatial separation between the X and Y electrodes is significant to make the electric field to propagate well in relation to the thickness of the dielectric panel.

A.2 Moisture Resistance

A useful side effect of the QMatrix acquisition method is that placing a floating conductive element between the X and Y lines tends to increase the field coupling and so increases the capacitance C_x . This is the opposite change direction to normal touch, and so can be quite easily ignored or compensated for by the controller. An example of such floating conductive elements is the water droplets caused by condensation.

As a result, QMatrix-based touchscreens tend not to go into false detect when they are covered in small non-coalesced water droplets. Once the droplets start to merge, however, they can become large enough to bridge the field across to nearby ground return paths (for example, other X lines not currently driven, or ground paths in mechanical chassis components). When this happens, the screen's behavior can become erratic.

There are some measures used in these controllers to help with this situation, but in general there comes a point where the screen is so contaminated by moisture that false detections become inevitable. It should also be noted that uniform condensation soon becomes non-uniform once a finger has spread it around. Finger grease renders the water highly conductive, making the situation worse overall.

In general, QMatrix has industry-leading moisture tolerance but there comes a point when even the best capacitive touchscreen suffers due to moisture on the dielectric surface.

-
1. A common problem with other types of capacitive acquisition technique when used for touchscreens, is that this isolation is not so pronounced. This means that when touching one region of the screen, the capacitive signals also tend to change slightly in nearby nodes too, causing small but often significant errors in the reported touch position.
 2. To a first approximation.
 3. Air gaps in front of QMatrix sensors massively reduce this field propagation and kill sensitivity. Normal optically clear adhesives work well to attach QMatrix touchscreens to their dielectric front panel.

A.3 Interference Sources

A.3.1 Power Supply

The device can tolerate short-term power supply fluctuations. If the power supply fluctuates slowly with temperature, the device tracks and compensates for these changes automatically with only minor changes in sensitivity. If the supply voltage drifts or shifts quickly, the drift compensation mechanism is not able to keep up, causing sensitivity anomalies or false detections.

If power supply noise is present (usually caused by LEDs, relays, or other high current devices) and affects the measurement results then a separate Low Dropout (LDO) type regulator should be used for the AVDD power supply.

It is recommended that all ceramic decoupling capacitors on supply lines are placed very close (<5 mm) to the chip. A bulk capacitor of at least 2.2 μF and a higher frequency capacitor of around 10 nF to 100 nF in parallel are recommended; both must be X7R or X5R dielectric capacitors.

A.3.2 Other Noise Sources

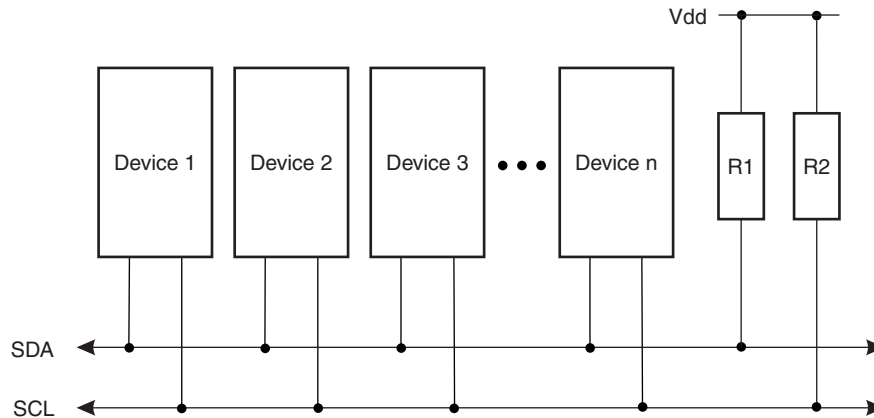
Refer to QTAN0079, *Buttons, Sliders and Wheels Sensor Design Guide*, for information (downloadable from the Touch Technology area of the Atmel website).

Appendix B. I²C Basics (I²C Operation)

B.1 Interface Bus

The device communicates with the host over an I²C bus. The following sections give an overview of the bus; more detailed information is available from www.nxp.com/documents/user_manual/UM10204.pdf. Devices are connected to the I²C bus as shown in Figure B-1. Both bus lines are connected to V_{dd} via pull-up resistors. The bus drivers of all I²C devices must be open-drain type. This implements a wired AND function that allows any and all devices to drive the bus, one at a time. A low level on the bus is generated when a device outputs a zero.

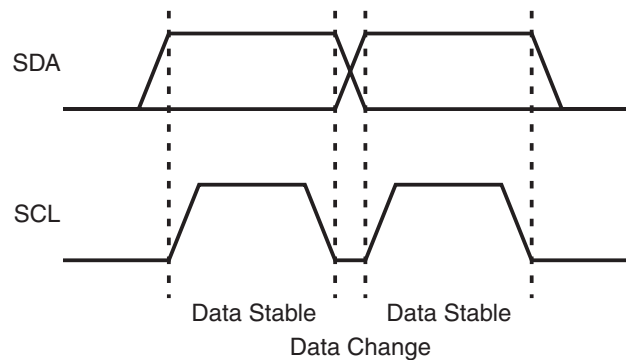
Figure B-1. I²C Interface Bus



B.2 Transferring Data Bits

Each data bit transferred on the bus is accompanied by a pulse on the clock line. The level of the data line must be stable when the clock line is high; the only exception to this rule is for generating START and STOP conditions.

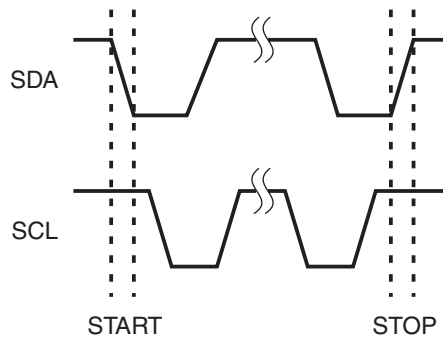
Figure B-2. Data Transfer



B.3 START and STOP Conditions

The host initiates and terminates a data transmission. The transmission is initiated when the host issues a START condition on the bus, and is terminated when the host issues a STOP condition. Between the START and STOP conditions, the bus is considered busy. As shown in [Figure B-3](#), START and STOP conditions are signaled by changing the level of the SDA line when the SCL line is high.

Figure B-3. START and STOP Conditions

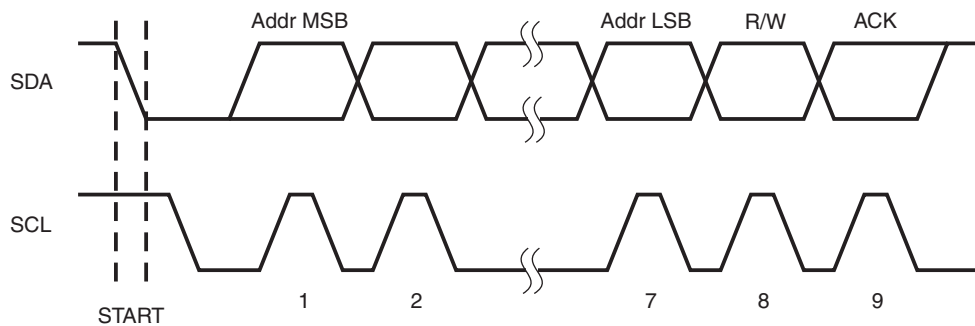


B.4 Address Byte Format

All address bytes are 9 bits long, consisting of 7 address bits, one READ/WRITE control bit and an acknowledge bit. If the READ/WRITE bit is set, a read operation is performed, otherwise a write operation is performed. When the device recognizes that it is being addressed, it will acknowledge by pulling SDA low in the ninth SCL (ACK) cycle. An address byte consisting of a slave address and a READ or a WRITE bit is called SLA+R or SLA+W, respectively.

The most significant bit of the address byte is transmitted first. The address sent by the host must be consistent with that selected with the option jumpers.

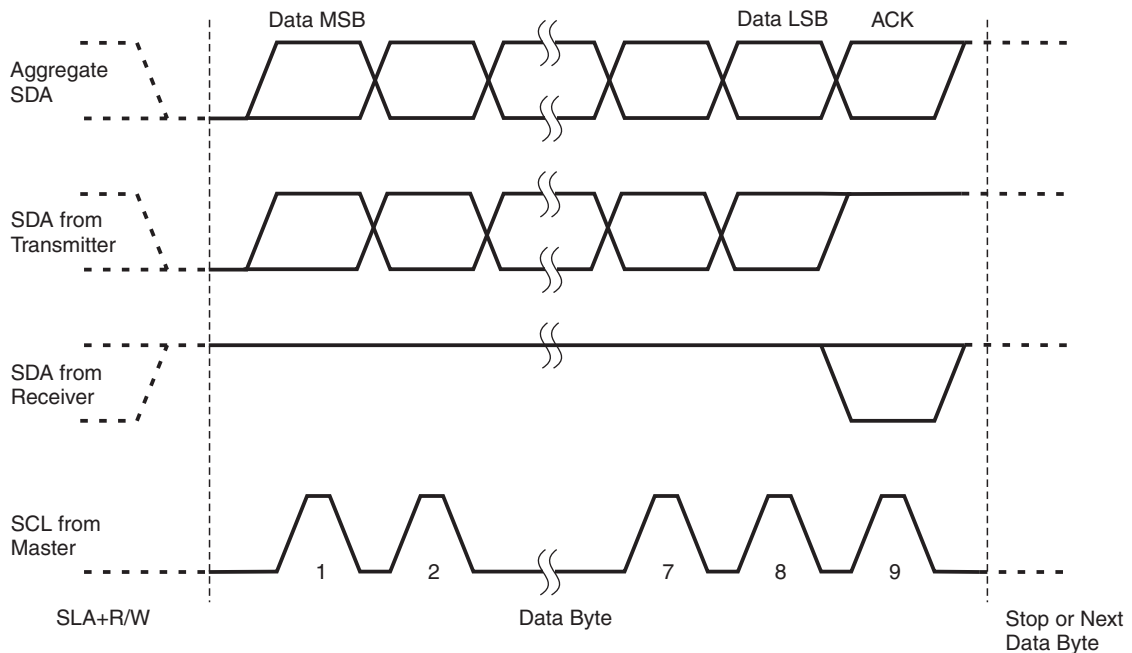
Figure B-4. Address Byte Format



B.5 Data Byte Format

All data bytes are 9 bits long, consisting of 8 data bits and an acknowledge bit. During a data transfer, the host generates the clock and the START and STOP conditions, while the receiver is responsible for acknowledging the reception. An acknowledge (ACK) is signaled by the receiver pulling the SDA line low during the ninth SCL cycle. If the receiver leaves the SDA line high, a NACK is signaled.

Figure B-5. Data Byte Format



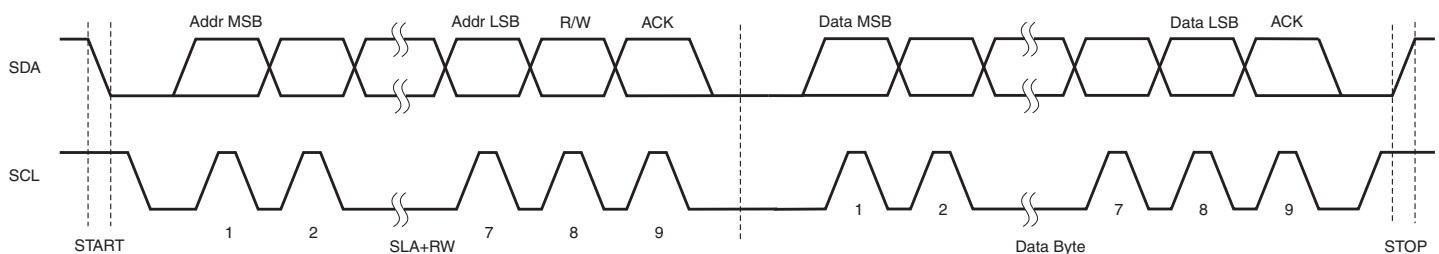
B.6 Combining Address and Data Bytes into a Transmission

A transmission consists of a START condition, an SLA+R/W, one or more data bytes and a STOP condition. The wired “ANDing” of the SCL line is used to implement handshaking between the host and the device. The device extends the SCL low period by pulling the SCL line low whenever it needs extra time for processing between the data transmissions.

Note: Each write or read cycle must end with a stop condition. The device may not respond correctly if a cycle is terminated by a new start condition.

Figure B-6 shows a typical data transmission. Note that several data bytes can be transmitted between the SLA+R/W and the STOP.

Figure B-6. Byte Transmission



Appendix C. Glossary of Terms

Channel

See **Node**.

Jitter

The peak-to-peak variance in the reported location for an axis when a fixed touch is applied. Typically jitter is random in nature and has a Gaussian ⁽¹⁾ distribution, therefore measurement of peak-to-peak jitter must be conducted over some period of time, typically a few seconds. Jitter is typically measured as a percentage of the axis in question.

For example a 100 × 100 mm Touchscreen that shows ±0.5% jitter in X and ±1% jitter in Y would show a peak deviation from the average reported coordinate of ±0.5 mm in X and ±1 mm in Y. Note that by defining the jitter relative to the average reported coordinate, the effects of linearity are ignored.

Linearity

The measurement of the peak-to-peak deviation of the reported touch coordinate in one axis relative to the absolute position of touch on that axis. This is often referred to as the nonlinearity. Non-linearity in either X or Y axes manifest themselves as regions where the perceived touch motion along that axis (alone) is not reflected correctly in the reported coordinate giving the sense of moving too fast or too slow. Linearity is measured as a percentage of the axis in question.

For each axis, a plot of the true coordinate versus the reported coordinate should be a perfect straight line at 45°. A non-linearity makes this plot deviate from this ideal line. It is possible to correct modest non-linearity using on-chip linearization tables, but this correction trades linearity for resolution in regions where stronger corrections are needed (because there is a stretching or compressing effect to correct the nonlinearity, so altering the resolution in these regions). Linearity is typically measured using data that has been sufficiently filtered to remove the effects of jitter. For example, a 100 mm slider with a nonlinearity of ±1% reports a position that is, at most, 1 mm away in either direction from the true position.

Multitouch

The ability of a touchscreen to report multiple concurrent touches. The touches are reported as separate sets of XY co-ordinates.

Node

One of the capacitive measurement points at which the sensor controller can detect capacitive change.

Resolution

The measure of the smallest movement on a slider or touchscreen in an axis that causes a change in the reported coordinate for that axis. Resolution is normally expressed in bits and tends to refer to resolution across the whole axis in question. For example, a resolution of 10 bits can resolve a movement of 0.0977 mm on a slider 100 mm long. Jitter in the reported position degrades usable resolution.

Touchscreen

A two-dimensional arrangement of electrodes whose capacitance changes when touched, allowing the location of touch to be computed in both X and Y axes. The output from the XY computation is a pair of numbers, typically 12-bits each, ranging from 0 to 4095, representing the extents of the touchscreen active region.

1. Sometimes called Bell-shaped or Normal distribution.

Associated Documents

Note: The documents listed below are available under NDA only. In addition, some documents may have further restrictions placed upon them.

For information on using and configuring the device, see the following:

- *mXT641T 1.0 Protocol Guide* (distributed with Atmel approval only)

The following documents may also be useful (available by contacting Atmel Touch Technology Division):

- **Touchscreen design and PCB/FPCB layout guidelines:**
 - Application Note: QTAN0054 – *Getting Started with maXTouch Touchscreen Designs*
 - Application Note: MXTAN0208 – *Design Guide for PCB Layouts for Atmel Touch Controllers*
 - Application Note: QTAN0080 – *Touchscreens Sensor Design Guide*
- **Configuring the device:**
 - Application Note: QTAN0078 – *maXTouch Stylus Tuning*
 - Application Note: QTAN0059 – *Using the maXTouch Self Test Feature*
 - Application Note: QTAN0070 – *Recovering from Palm Touches During Calibration with maXTouch Touchscreen Controllers*
 - Application Note: MXT0202 – *Using the Unlock Gesture T81 Object*
- **Miscellaneous:**
 - Application Note: QTAN0050 – *Using the maXTouch Debug Port*
 - Application Note: QTAN0058 – *Rejecting Unintentional Touches with the maXTouch Touchscreen Controllers*
 - Application Note: QTAN0061 – *maXTouch Sensitivity Effects for Mobile Devices*
 - Application Note: QTAN0083 – *Power and Speed Considerations*
 - Application Note: QTAN0051 – *Bootloading Procedure for Atmel Touch Sensors Based on the Object Protocol*
- **Tools:**
 - QTAN0101 – *Object Server User Guide*

Revision History

Revision Number	History
Revision ASX – September 2013	Initial release – Summary
Revision BSX – October 2013	<ul style="list-style-type: none"> • Front page Features updated • Section 1 (Overview): minor updates • Section 2.3 (Schematic): updated to reduce bypass capacitors and simplify VDDCORE circuit. Schematics notes section added.
Revision CSX – October 2013	<ul style="list-style-type: none"> • Front page Features updated: I²C-compatible interface information added
Revision DSX – October 2013	<ul style="list-style-type: none"> • Updated pinout table. G9 correctly identified as X8
Revision ESX – November 2013	<ul style="list-style-type: none"> • Updated pin configuration diagram to show pins from top side • Schematic updated with additional notes • Avdd specification updated
Revision FSX – December 2013	<ul style="list-style-type: none"> • XVdd voltage doubler correctly identified as connected to Vdd and not AVdd • VDDIN renamed as VDD
Revision GX – March 2014	<ul style="list-style-type: none"> • Additional sections added • Updated to firmware 1.0 – Preliminary
Revision HX – May 2014	<ul style="list-style-type: none"> • Revised – Release
Revision IX – January 2015	<ul style="list-style-type: none"> • Revised voltage regulator recommendations
Revision JX – February 2015	<ul style="list-style-type: none"> • Updated Power-up / Reset Requirements section

Notes



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